

Time Response

- The mathematical representation of a system (*Transfer function or State space*) is used to analyze its transient and steady-state responses to see if these characteristics yield the desired behavior.
- This chapter is devoted to the analysis of *system transient response*.

Time Response

Poles, Zeros, and System Response

- The output response of a system is the sum of two responses:
 - the forced response* (steady-state response or particular solution),
 - the natural response* (the homogeneous solution).

Output response = forced response (e.g. constant) + natural response (e.g. exponential)

- Poles** of a Transfer Function (TF): The values of **s** that cause $TF \rightarrow \infty$

roots of the denominator (characteristic polynomial) of the transfer function

- Zeros** of a TF: the values of **s** that cause $TF = 0$.

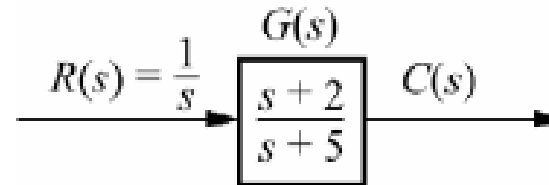
roots of the numerator of the transfer function

Example: Poles and Zeros of a First-Order System

$$C(s) = \frac{(s+2)}{s(s+5)} = \frac{A}{s} + \frac{B}{s+5} = \frac{2/5}{s} + \frac{3/5}{s+5}$$

System Output (unit step response)

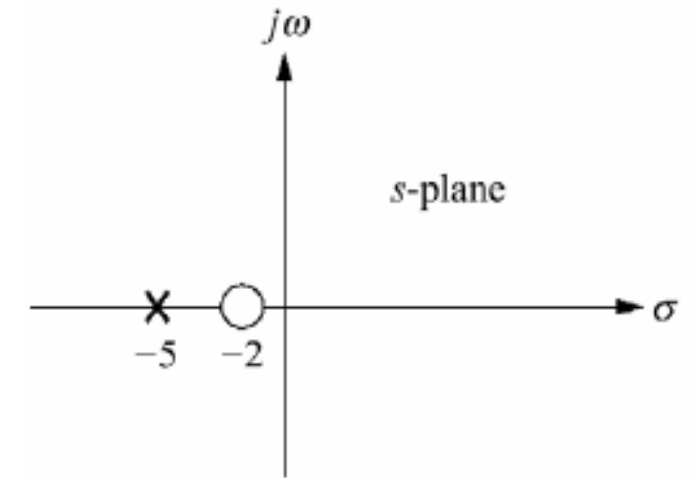
where $A = \frac{(s+2)}{(s+5)} \Big|_{s=0} = \frac{2}{5}$ and $B = \frac{(s+2)}{s} \Big|_{s=-5} = \frac{3}{5}$



System showing input and output

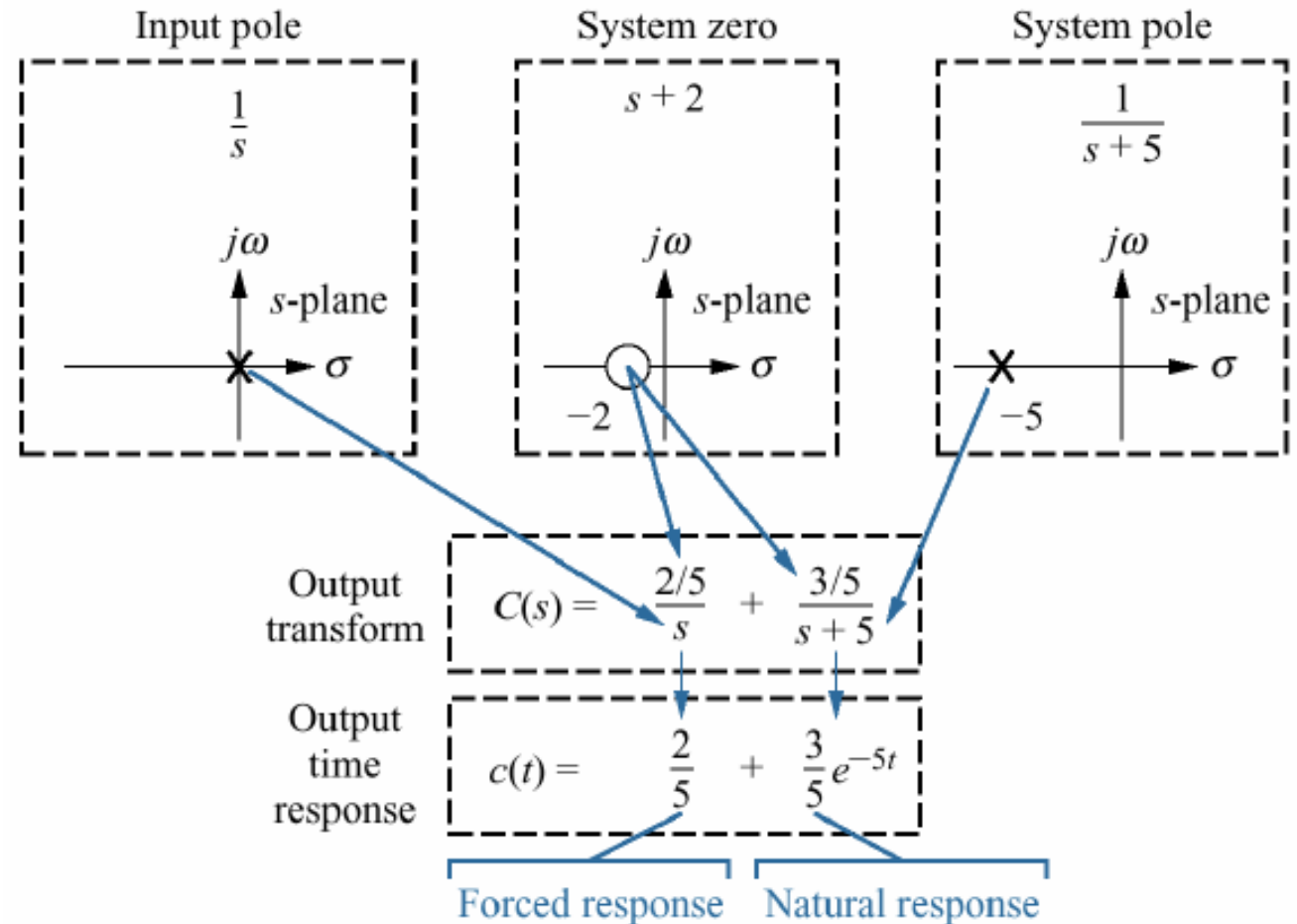
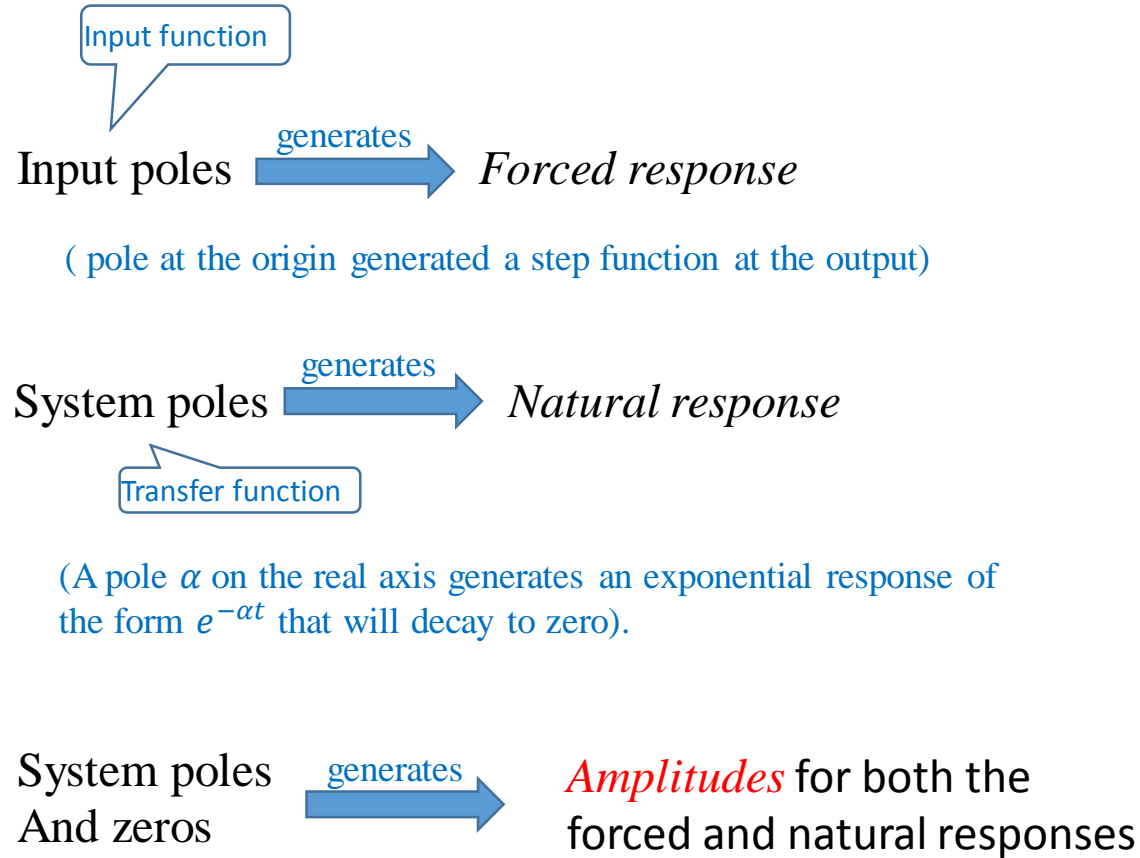
Inverse Laplace transform: $c(t) = \frac{2}{5} + \frac{3}{5}e^{-5t}$

System Output in time domain (time response)



pole-zero plot of the system

Example: Poles and Zeros of First Order System

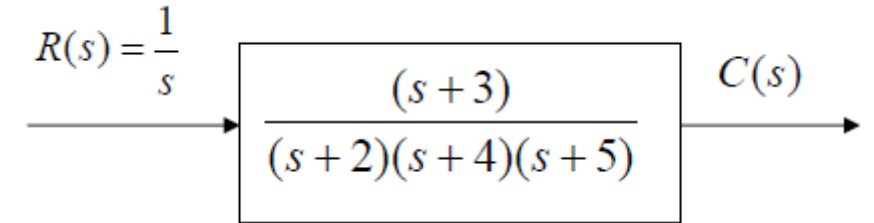


Evolution of a system response.

Evaluating Response using Poles

Problem:

Given the following system, write the output, $c(t)$, in general terms. Specify the forced and natural parts of the solution.



Solution:

$$C(s) = \underbrace{\frac{K_1}{s}}_{\text{Forced response}} + \underbrace{\frac{K_2}{(s+2)} + \frac{K_3}{(s+4)} + \frac{K_4}{(s+5)}}_{\text{Natural response}}$$

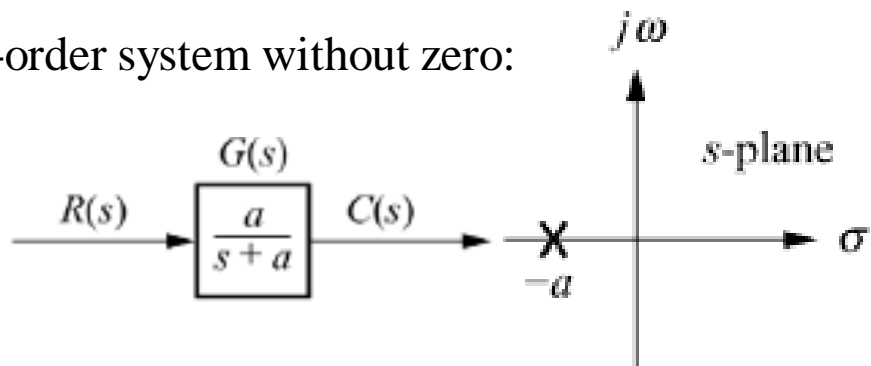
Taking inverse Laplace transform,

- Each system pole generates an exponential as part of the natural response.
- The input's pole generates the forced response.

$$c(t) = \underbrace{K_1}_{\text{Forced response}} + \underbrace{K_2 e^{-2t} + K_3 e^{-4t} + K_4 e^{-5t}}_{\text{Natural response}}$$

First-Order System: Time Constant

- A first-order system without zero:



- If the input is a unit step: $R(s) = \frac{1}{s}$ then the Laplace transform of the step response is :

$$C(s) = R(s)G(s) = \frac{a}{s(s+a)}$$

the input pole at the origin generated the forced response

- Taking the inverse transform

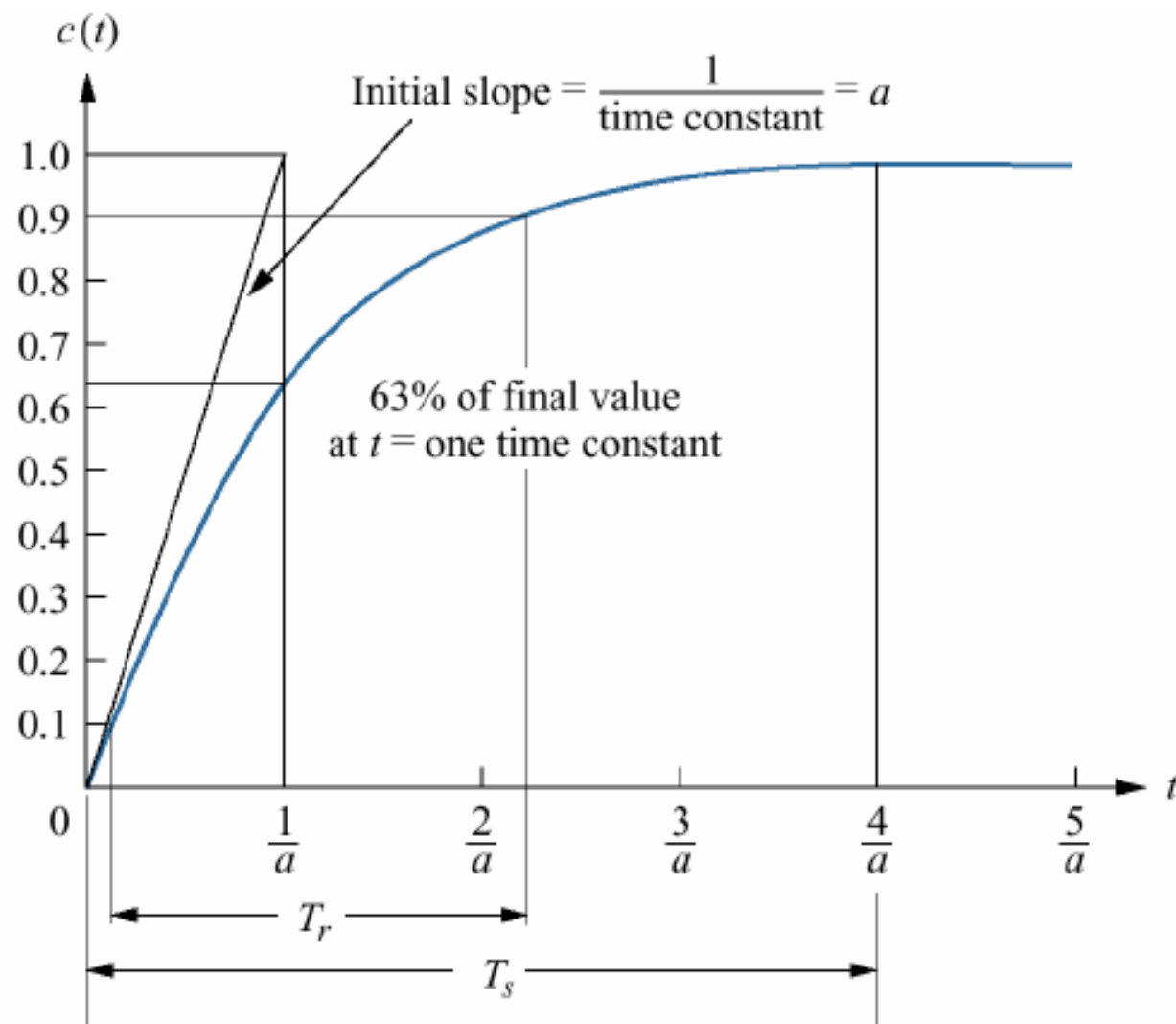
$$c(t) = c_f(t) + c_n(t) = 1 - e^{-at}$$

The system pole at $-a$ generated the natural response

- Significance of parameter a (only parameter needed to describe the transient response),

When $t = 1/a$, $e^{-at} \Big|_{t=1/a} = e^{-1} = 0.37$

Hence, $c(t) \Big|_{t=1/a} = 1 - 0.37 = 0.63$



Some Terminology

(three transient response performance specifications).

1. **Time constant T_c** : Time it takes for the step response to rise to 63% of its final value.

- we can call the parameter a (*system pole*) the *exponential frequency* (The reciprocal of the time constant)
- T_c is related to the speed at which the system responds to a step input.

$$T_c = \frac{1}{a}$$

2. **Rise Time T_r** : Rise time is defined as the time for the response to go from 0.1 to 0.9 of its final value.

found by solving for the difference in time at $c(t) = 0.9$ and $c(t) = 0.1$

$$\begin{aligned} C(t_2) = 0.9 &= 1 - e^{-at_2} \Rightarrow t_2 = -\frac{\ln(0.1)}{a} = \frac{2.31}{a} \\ C(t_1) = 0.1 &= 1 - e^{-at_1} \Rightarrow t_1 = -\frac{\ln(0.9)}{a} = \frac{0.11}{a} \end{aligned} \Rightarrow T_r = t_2 - t_1 = \frac{2.13}{a} - \frac{0.11}{a} = \frac{2.2}{a}$$

time for $C(t) = 0.1$
time for $C(t) = 0.9$

Rise time: $T_r = \frac{2.2}{a}$

3. **Settling time T_s** : The time for the response to reach, and stay within, 2% of its final value.

Letting $C(t) = 0.98$ and solving for time, t , we find the settling time to be

$$C(T_s) = 0.98 = 1 - e^{-at_2} \Rightarrow T_s = -\frac{\ln(0.98)}{a} = \frac{4}{a}$$

$$T_s = \frac{4}{a}$$

First-Order Transfer Function via Testing

- With a step input, we can measure the time constant and the steady-state value, from which the transfer function can be calculated.
- A simple first order system has : $G(s) = K / (s+a)$,
and step response is: $C(s) = \frac{K}{s(s+a)} = \frac{K/a}{s} - \frac{K/a}{(s+a)}$
- From *the response*, we identify K and a to obtain the transfer function.

To find a :

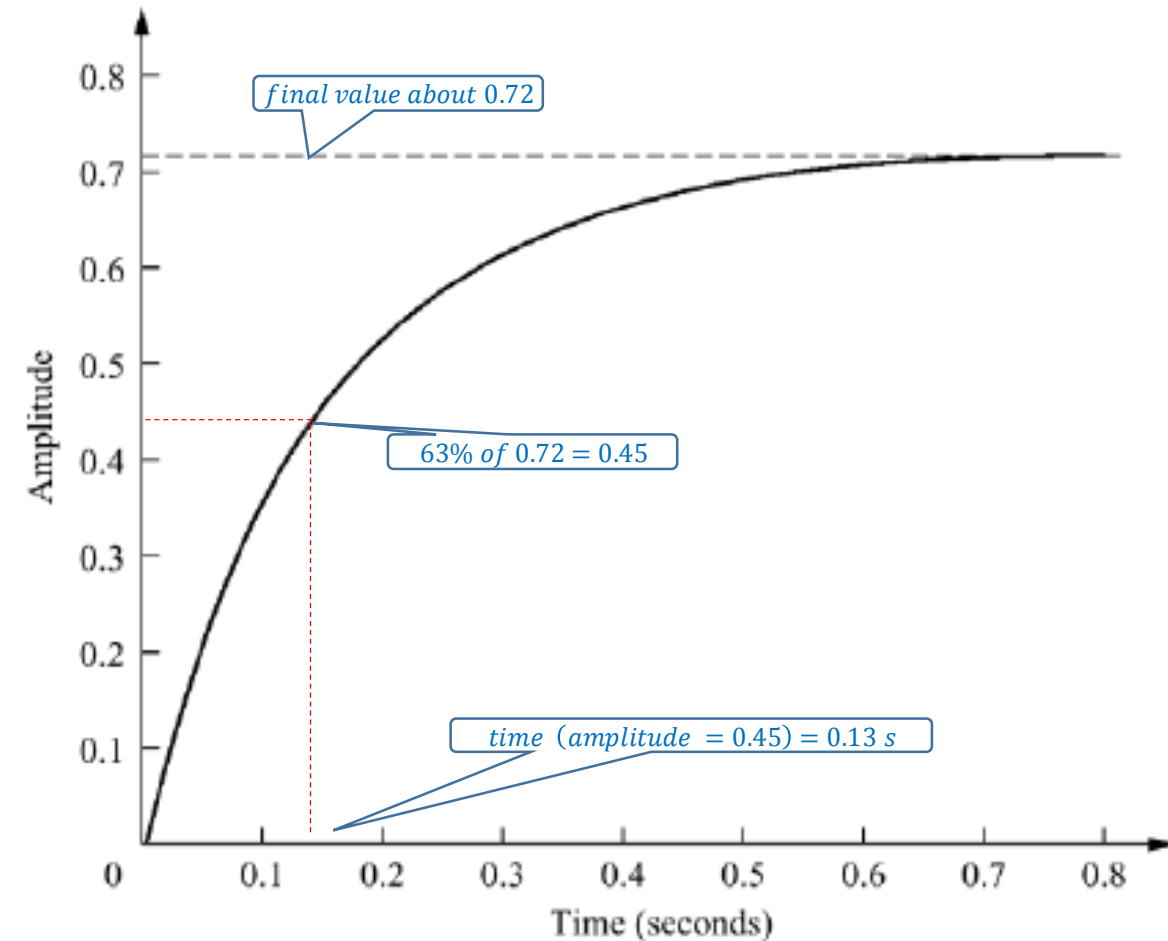
$$\begin{aligned} \text{Time constant} &= \text{Time}(0.63 \times 0.72) \\ &= \text{Time}(0.45) = 0.13 \text{ second} \end{aligned}$$

$$a = 1/0.13 = 7.7$$

To find K : The forced response reaches a steady-state value of $K/a = 0.72$ → $K = 5.54$

Transfer function,

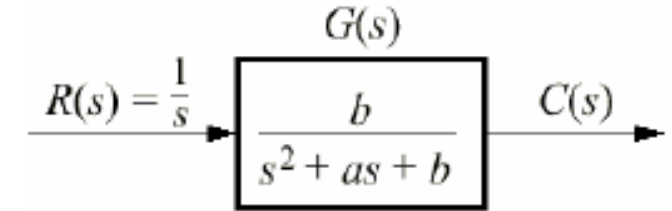
$$G(s) = \frac{5.54}{(s + 7.7)}$$



Second-Order System

- Parameters of First-order system determine the *speed* of the system.
- Parameters of Second-order system determine the *form (shape)* of the system.
- Some examples of second-order response:

By assigning appropriate values to parameters a and b , we can show all possible second-order transient responses.

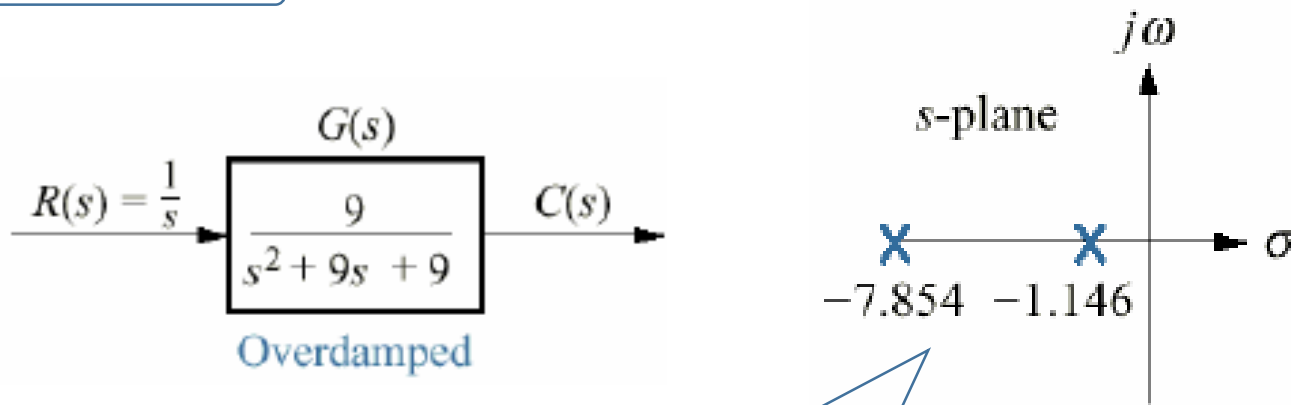


the general case
(two finite poles and no zeros)

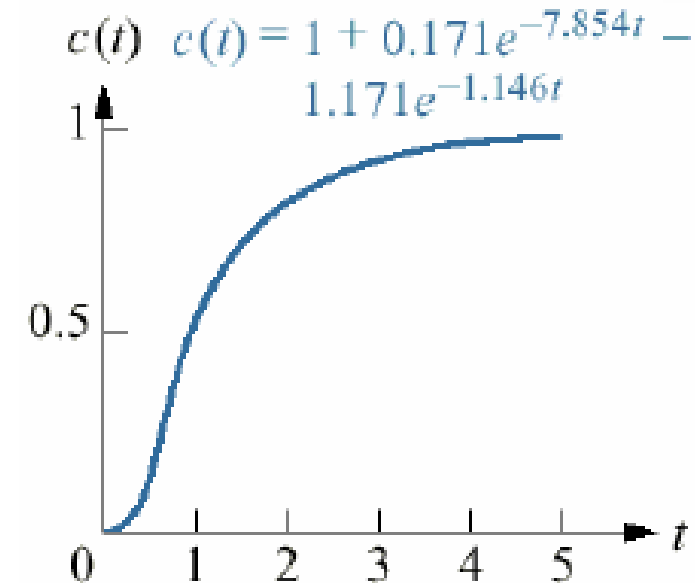
- Over-damped Response:** (pole at the origin from the unit step input and Two real poles that come from the system)

$$C(s) = \frac{9}{s(s^2 + 9s + 9)} = \frac{9}{s(s + 7.854)(s + 1.146)}$$

$a = 9$ and $b = 9$



Two real poles

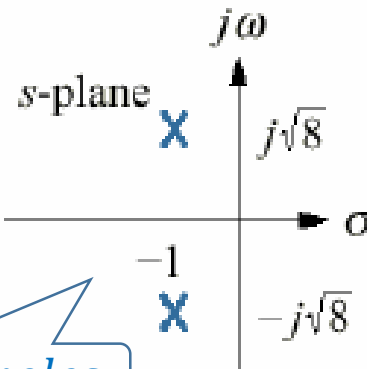
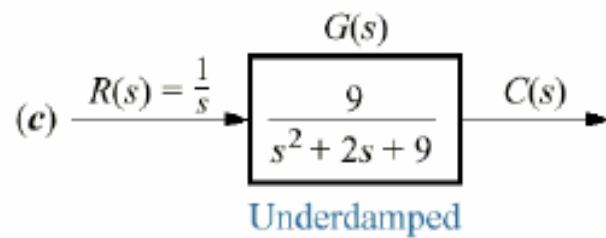


Second-Order System

2. Under-damped Response: (pole at the origin from the input and Two complex poles that come from the system).

$$C(s) = \frac{9}{s(s^2 + 2s + 9)} = \frac{9}{s(s + [1 + j\sqrt{8}])(s + [1 - j\sqrt{8}])}$$

$a = 2$ and $b = 9$



Two complex poles

$$c(t) = K_1 + e^{-t}(K_2 \cos \sqrt{8}t + K_3 \sin \sqrt{8}t)$$

$$= K_1 + K_4 e^{-t}(\cos \sqrt{8}t - \varphi)$$

Where,

$$\varphi = \tan^{-1} \frac{K_3}{K_2}, \text{ and } K_4 = \sqrt{K_2^2 + K_3^2}$$

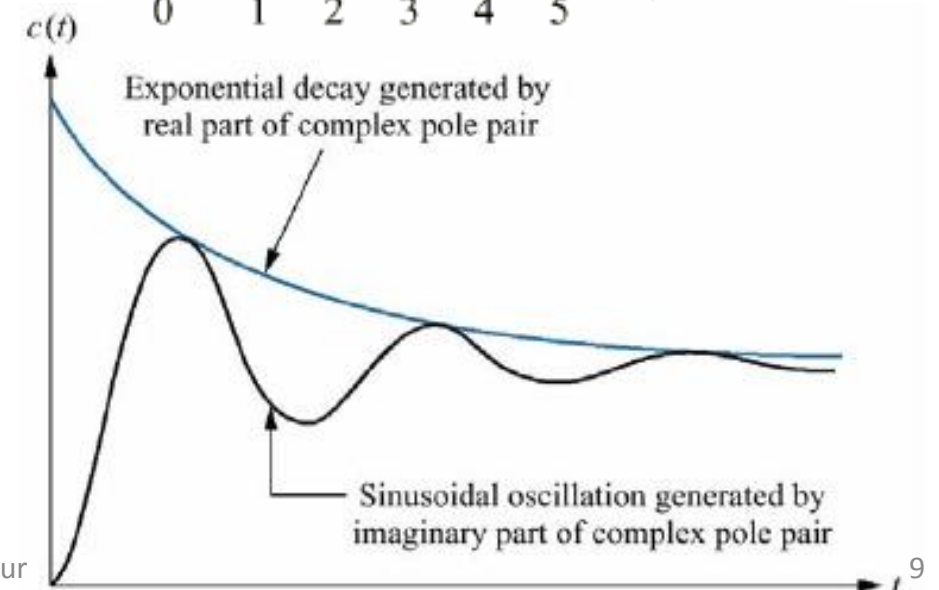
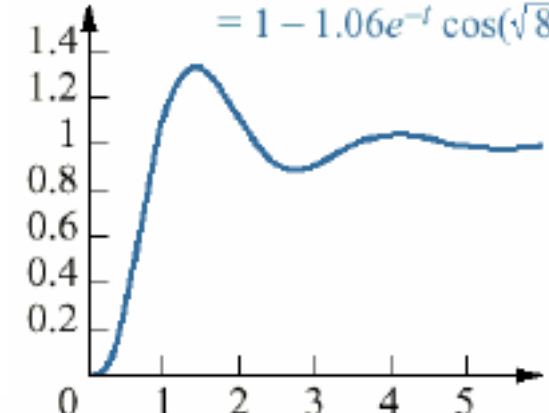
Poles from the system: $s = -1 \pm j\sqrt{8}$

exponential decay

frequency of the sinusoidal oscillation.

$$c(t) = 1 - e^{-t}(\cos \sqrt{8}t + \frac{\sqrt{8}}{8} \sin \sqrt{8}t)$$

$$= 1 - 1.06e^{-t} \cos(\sqrt{8}t - 19.47^\circ)$$



Second-order
step response
components
generated by
complex poles

Second-Order System

3. Un-damped Response: pole at the origin that comes from the input and two imaginary poles that come from the system.

$$C(s) = \frac{9}{s(s^2 + 9)}$$

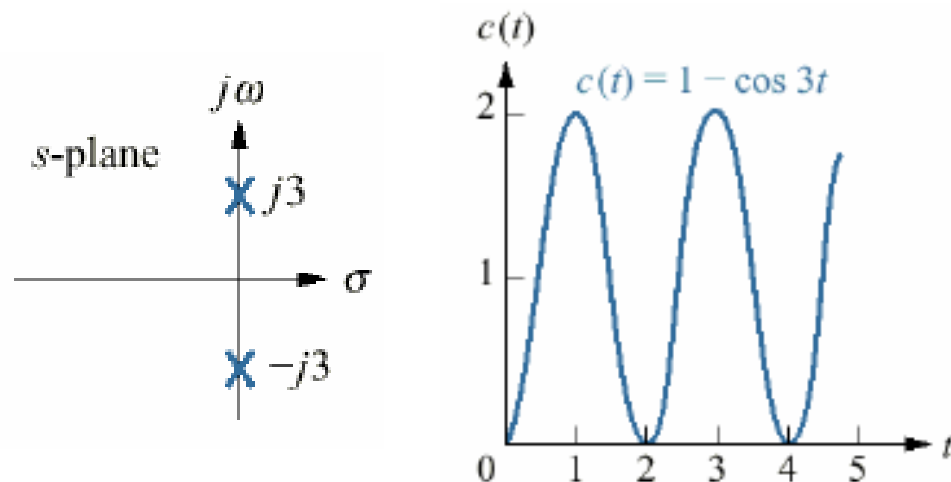
$a = 0$ and $b = 9$

$$s = \pm j3$$

two system poles on the imaginary axis

$$c(t) = K_1 + K_4 \cos(3t - \varphi)$$

There is no exponential term, so no decay.



4. Critically Damped Response: pole at the origin that comes from the input and two multiple real poles that come from the system.

$$C(s) = \frac{9}{s(s^2 + 6s + 9)} = \frac{9}{s(s+3)^2}$$

$a = 6$ and $b = 9$

$$s = -3, -3$$

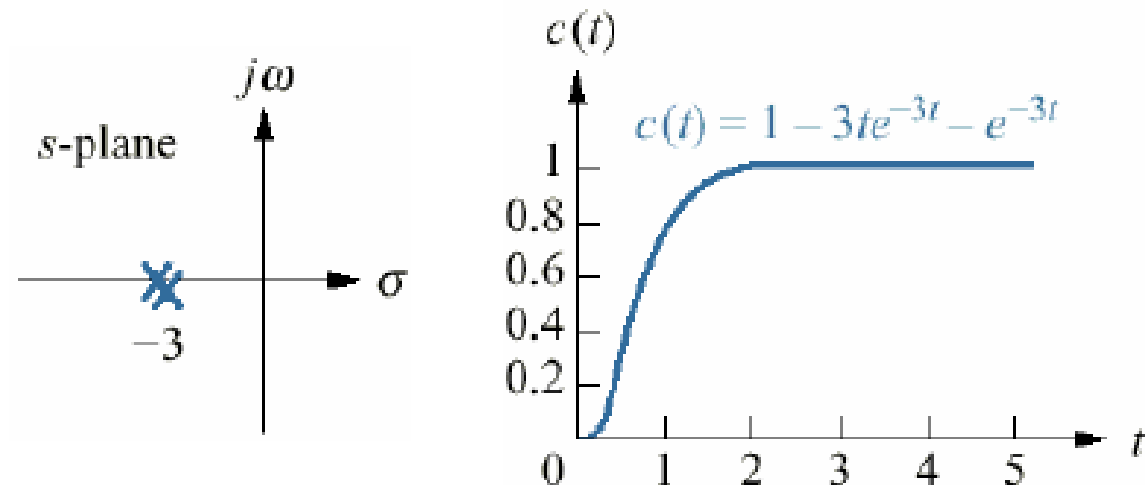
two poles (double) on the real axis at -3

input pole

system poles

$$c(t) = K_1 + K_2 te^{-3t} + K_3 te^{-3t}$$

There is no sinusoidal term, so no oscillation.



Second-Order System

All Together

Over-damped responses

Two real poles at $-\sigma_1, -\sigma_2$

$$c(t) = K_1 e^{-\sigma_1 t} + K_2 e^{-\sigma_2 t}$$

Under-damped responses

Two complex poles at $-\sigma_d \mp j\omega_d$

$$c(t) = A e^{-\sigma_d t} \cos(\omega_d t - \phi)$$

Un-damped responses

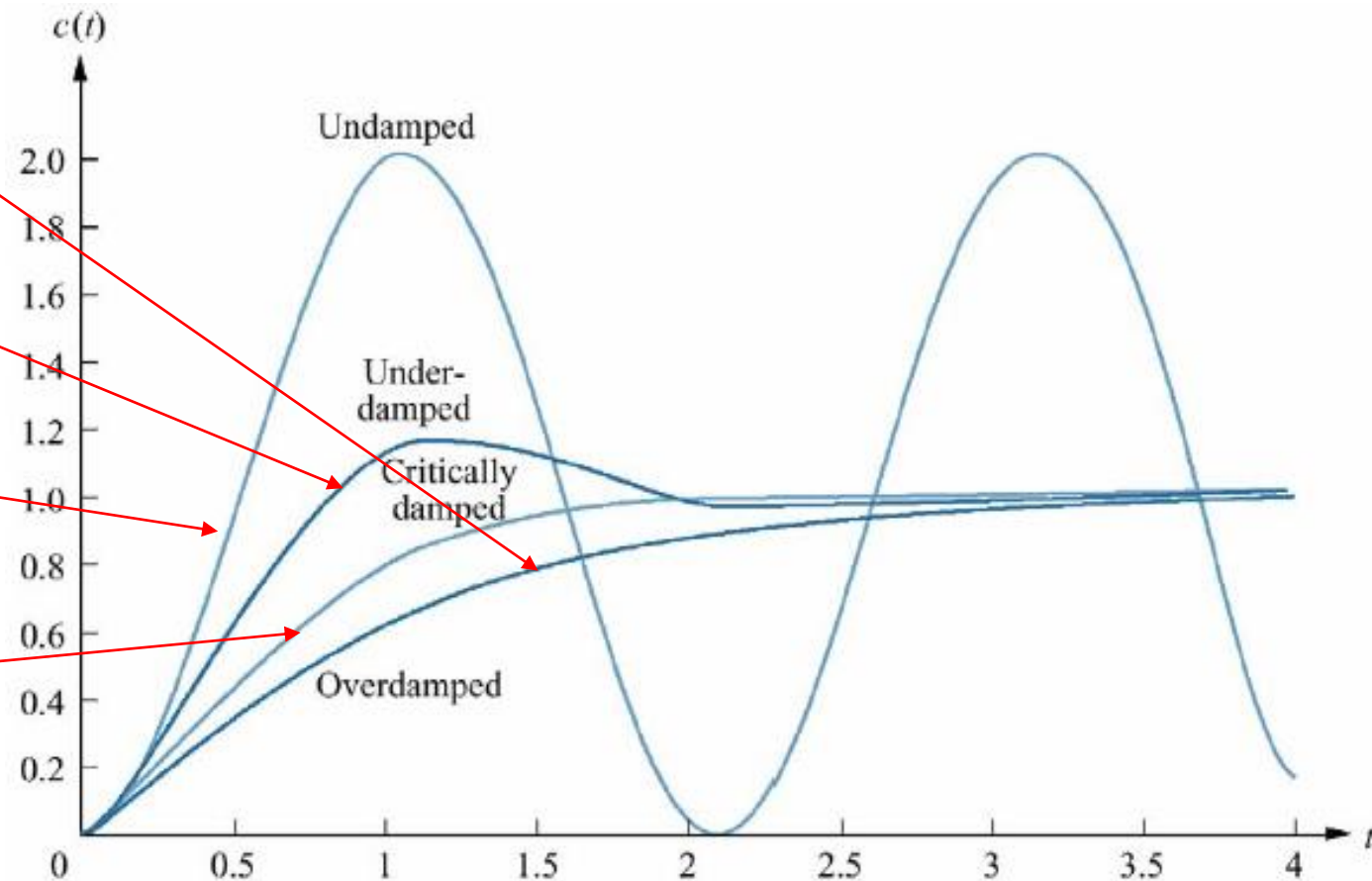
Two imaginary poles at $\mp j\omega_1$

$$c(t) = A \cos(\omega_1 t - \phi)$$

Critically damped responses

Two real poles at $-\sigma_1$

$$c(t) = K_1 e^{-\sigma_1 t} + K_2 t e^{-\sigma_1 t}$$



Step responses for second-order system damping cases

The General Second-Order System

- we define two physically meaningful specifications to describe the characteristics of the second-order transient response systems, *natural frequency* and *damping ratio*.

Natural Frequency ω_n : the frequency of oscillation of the system without damping.

Damping Ratio ξ : dimensionless measure describing how oscillations in a system decay.

$$\xi = \frac{\text{Exponential Decay Frequency}}{\text{Natural Frequency (rad/second)}} = \frac{1}{2\pi} \frac{\text{Natural period (seconds)}}{\text{Exponential time constant}}$$

- Generalization of second-order system in terms of ξ and ω_n .

Consider the general system, $G(s) = \frac{b}{s^2 + as + b}$

For un-damped(without damping) system, $a = 0$, and the poles are on $j\omega - axis$ at $\mp j\sqrt{b}$, $\longleftarrow G(s) = \frac{b}{s^2 + b}$

$$\therefore \omega_n = \sqrt{b}, \text{ Hence } b = \omega_n^2$$

For an under-damped system, poles have real part $\sigma = -a/2$ (exponential decay)

$$\xi = \frac{|\sigma|}{\omega_n} = \frac{a/2}{\omega_n} \Rightarrow a = 2\xi\omega_n$$

Exponential decay frequency

Natural frequency

The general second-order transfer function finally looks like

$$G(s) = \frac{\omega_n^2}{s^2 + 2\xi\omega_n s + \omega_n^2}$$

Second-Order System As a Function of Damping Ratio

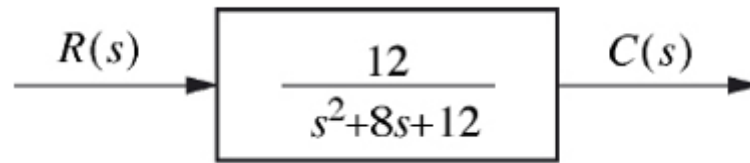
- Relationship between the quantities ω_n and ξ and the pole location.

Solving for the poles of the transfer function

$$G(s) = \frac{\omega_n^2}{s^2 + 2\xi\omega_n s + \omega_n^2} \xrightarrow{\text{Denominator roots}} s_{1,2} = -\xi\omega_n \pm \omega_n\sqrt{\xi^2 - 1}$$

Example

For the system find the value of ξ and report the kind of response expected.

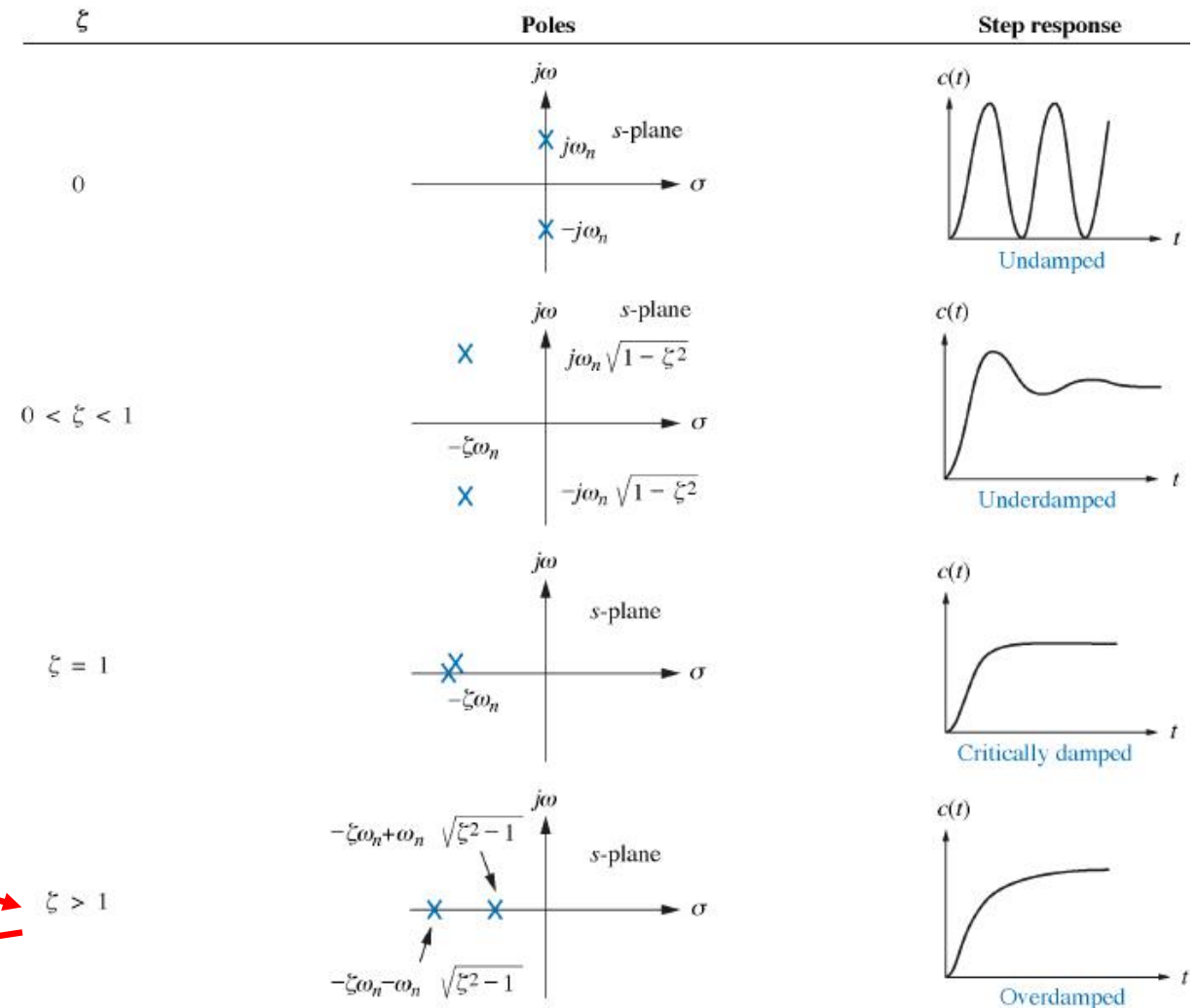


We have $G(s) = \frac{\omega_n^2}{s^2 + 2\xi\omega_n s + \omega_n^2} = \frac{12}{s^2 + 8s + 12}$

$\Rightarrow \omega_n^2 = 12 \Rightarrow \omega_n = \sqrt{12}$

and $2\xi\omega_n = 8 \Rightarrow \xi = \frac{8}{2\sqrt{12}} = \frac{2}{\sqrt{3}} > 1$

System is over-damped.



Underdamped Second-Order Systems

- The nature of the response obtained is related to the value of the damping ratio ξ (over-damped, critically damped, underdamped, and un-damped responses.).
- Step response for the general second-order system,

$$C(s) = R(s)G(s) = \frac{1}{s} \frac{\omega_n^2}{s^2 + 2\xi\omega_n s + \omega_n^2} \rightarrow C(s) = \frac{\omega_n^2}{s(s^2 + 2\xi\omega_n s + \omega_n^2)} = \frac{K_1}{s} + \frac{K_2 s + K_3}{s^2 + 2\xi\omega_n s + \omega_n^2}$$

Expanding by partial fractions, ($\xi < 1$ the underdamped case)

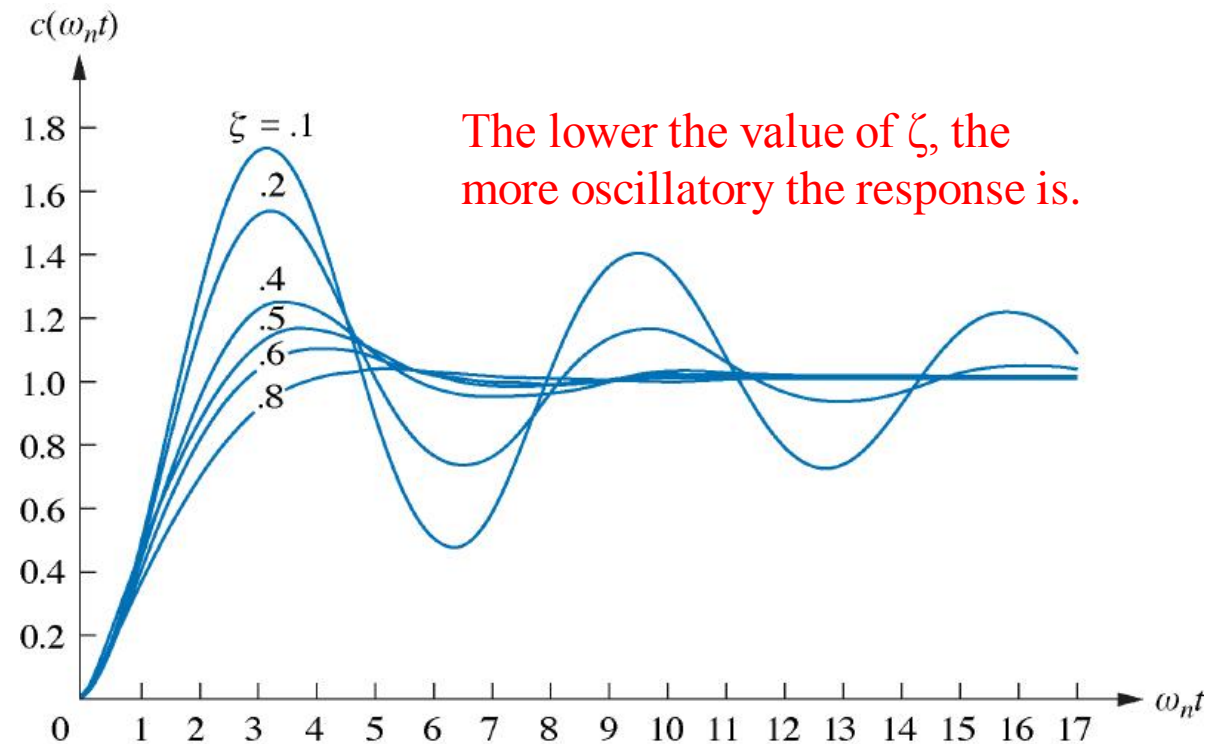
$$C(s) = \frac{1}{s} - \frac{(s + \xi\omega_n) + \frac{\xi}{\sqrt{1-\xi^2}}\omega_n\sqrt{1-\xi^2}}{(s + \xi\omega_n)^2 + \omega_n^2(1-\xi^2)}$$

inverse
Laplace
transform

$$\rightarrow c(t) = 1 - e^{-\xi\omega_n t} \left(\cos \omega_n \sqrt{1-\xi^2} t + \frac{\xi}{\sqrt{1-\xi^2}} \sin \omega_n \sqrt{1-\xi^2} t \right)$$

$$= 1 - \frac{1}{\sqrt{1-\xi^2}} e^{-\xi\omega_n t} \cos(\omega_n \sqrt{1-\xi^2} t - \varphi)$$

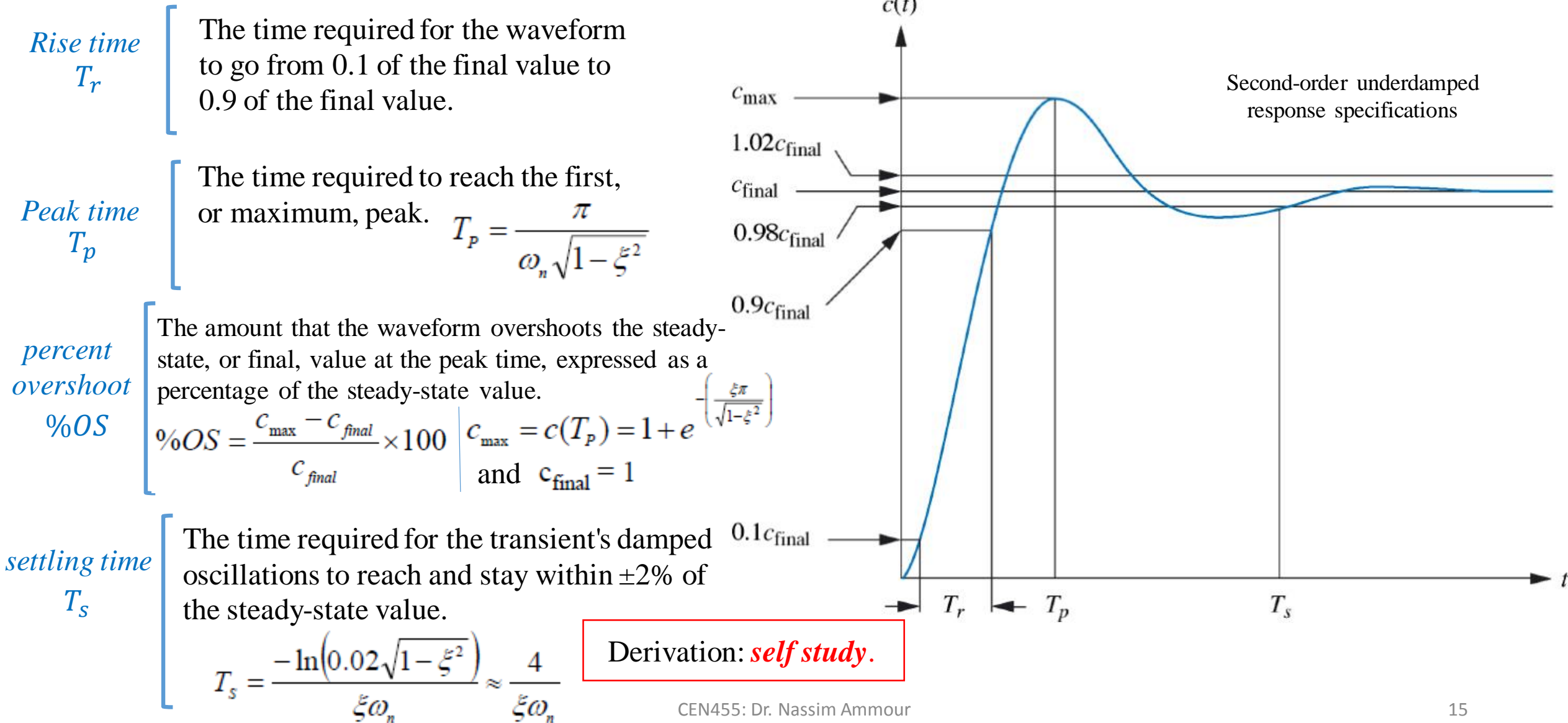
Where, $\varphi = \tan^{-1} \left(\frac{\xi}{\sqrt{1-\xi^2}} \right)$



Underdamped Second-Order Systems

Specifications

- Other parameters associated with the underdamped response are *rise time*, *peak time*, *percent overshoot*, and *settling time*.



Under-damped Second-Order Systems

Specifications (continued)

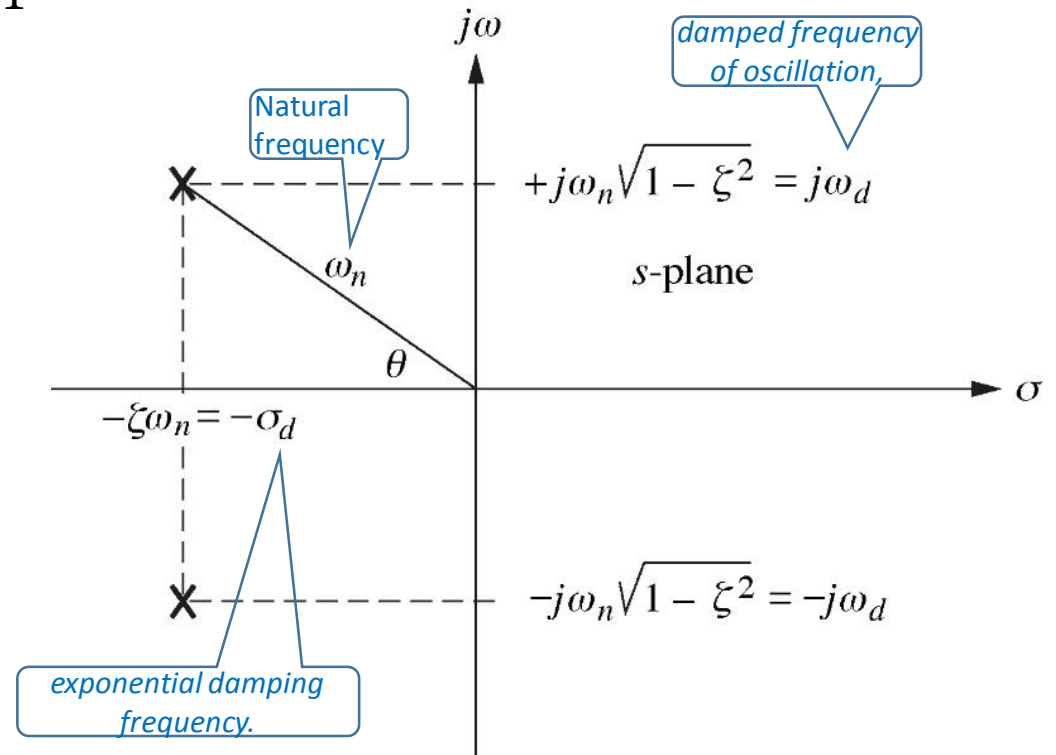
from the Pythagorean theorem

$$\left. \begin{array}{l} \omega_d^2 + \sigma_d^2 = \omega_n^2 \\ \sigma_d = \xi \omega_n \end{array} \right\} \Rightarrow \omega_d = \sqrt{\omega_n^2 - \sigma_d^2} = \sqrt{\omega_n^2 - \xi^2 \omega_n^2} = \omega_n \sqrt{1 - \xi^2}$$

$$\cos(\theta) = \frac{\sigma_d}{\omega_n} = \frac{\xi \omega_n}{\omega_n} = \xi$$

$$T_P = \frac{\pi}{\omega_n \sqrt{1 - \xi^2}} \Rightarrow T_P = \frac{\pi}{\omega_n \sqrt{1 - \xi^2}} = \frac{\pi}{\omega_d}$$

$$T_s = \frac{-\ln(0.02 \sqrt{1 - \xi^2})}{\xi \omega_n} \approx \frac{4}{\xi \omega_n} \Rightarrow T_s = \frac{4}{\xi \omega_n} = \frac{4}{\sigma_d}$$



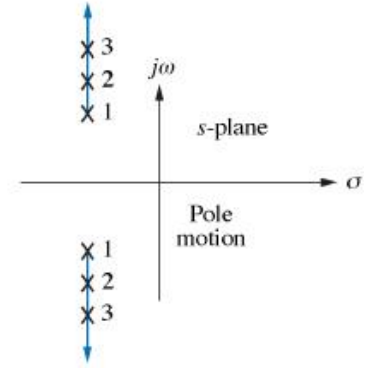
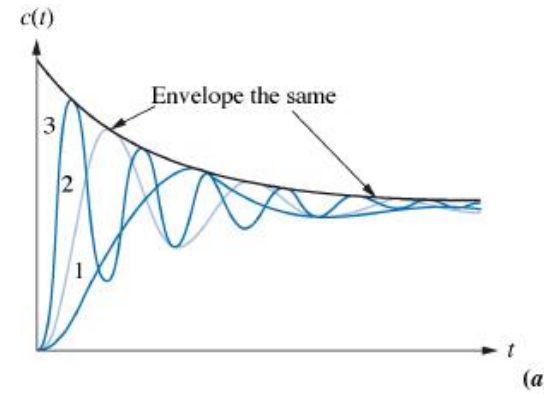
Under-damped Second-Order Systems

Step Response as Pole moves

poles move in a vertical direction (with constant real part)



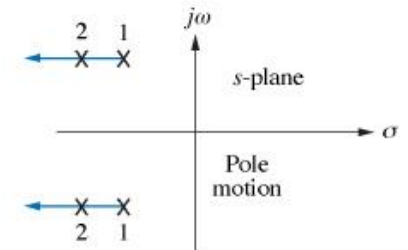
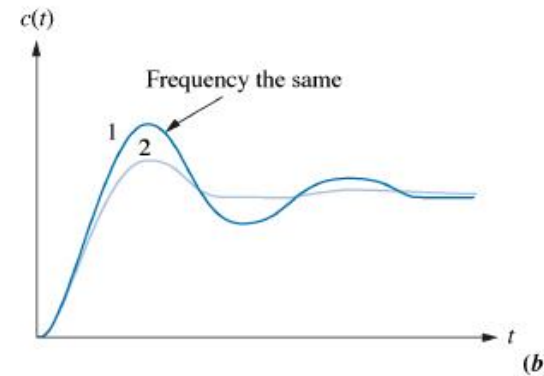
- frequency increases
- envelope remains the same (constant real part)
- settling time is virtually the same
- overshoot increases, the rise time decreases



poles move in a horizontal direction (with constant imaginary part)



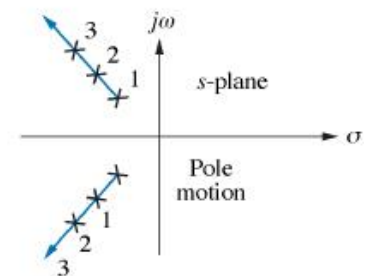
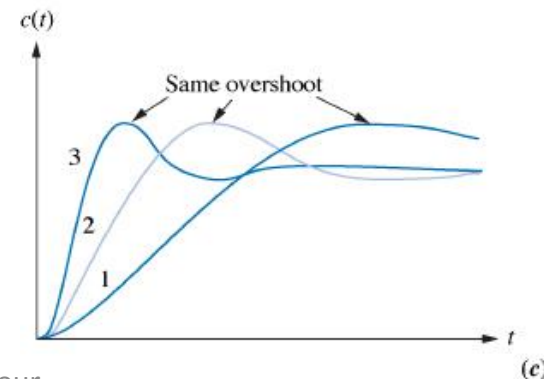
- As the poles move to the left, response damps out more rapidly.
- peak time is the same for all waveforms (constant imaginary part)



poles move in along a constant radial line direction



- The percent overshoot remains the same.
- The farther the poles are from the origin, the more rapid the response.



Finding TP, %OS, and TS From Pole Location

Problem: Given the pole plot find ξ , ω_n , T_p , %OS, and T_s .

Solution: _____

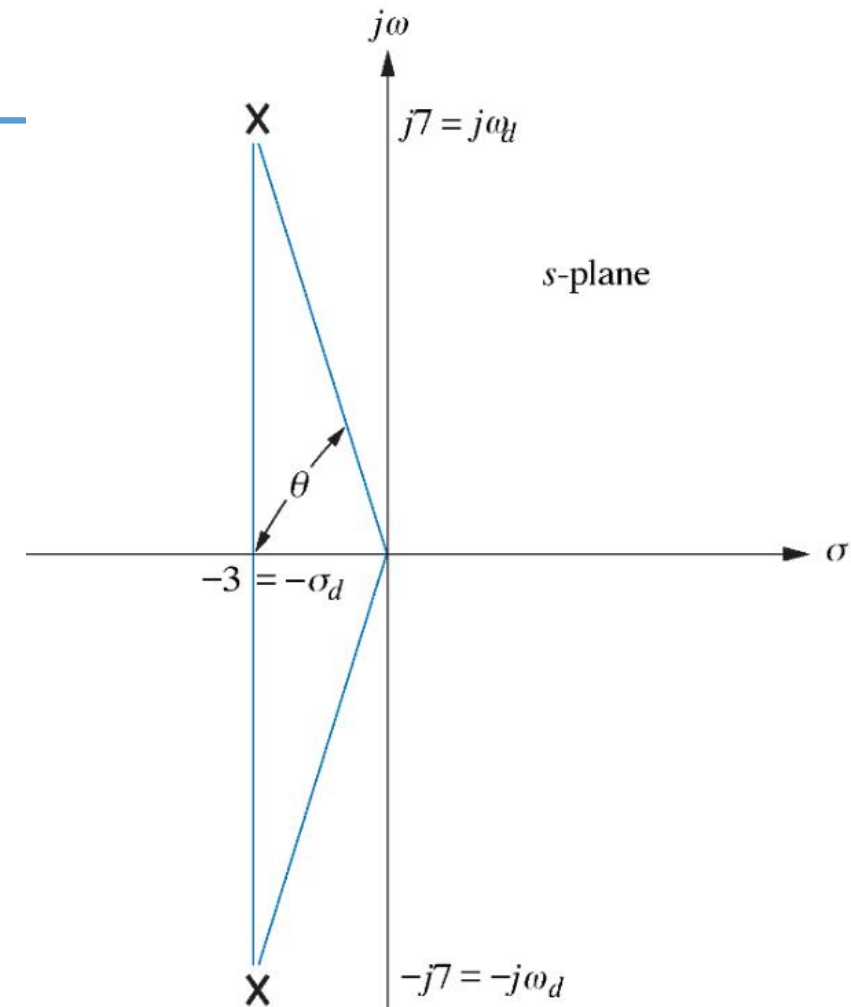
Damping ration, $\xi = \cos(\theta) = \cos[\arctan(7 / 3)] = 0.394$.

Natural frequency, $\omega_n = \sqrt{\omega_d^2 + \sigma_d^2} = \sqrt{7^2 + 3^2} = 7.616$

Peak time, $T_p = \frac{\pi}{\omega_d} = \frac{\pi}{7} = 0.449 \text{ second}$

Percent overshoot, $\%OS = e^{-\left(\xi\pi / \sqrt{1-\xi^2}\right)} \times 100 = 26\%$

The approximate settling time, $T_s = \frac{4}{\sigma_d} = \frac{4}{3} = 1.333 \text{ second}$



System Response with Additional Poles

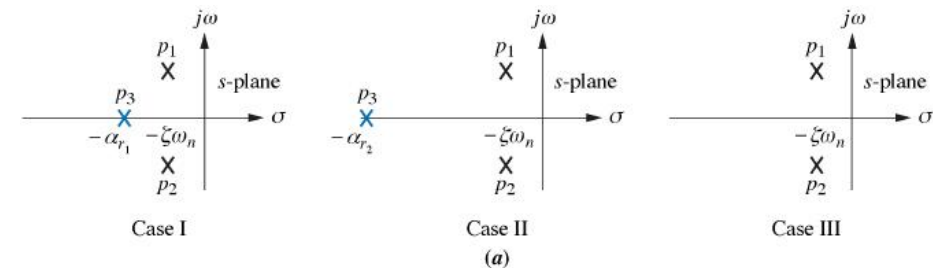
- If a system has more than two poles or has zeros, we cannot use the formulas to calculate the performance specifications that we derived.
- We need to approximate that system to a second-order system that has just two dominant complex poles.

Assuming two complex poles at $-\xi\omega_n \pm j\omega_n\sqrt{1-\xi^2}$ And the real pole at $-\alpha_r$

Time domain step response, $c(t) = Au(t) + e^{-\xi\omega_n t}(B \cos \omega_d t + C \sin \omega_d t) + De^{-\alpha_r t}$

Case I: α_r is not much larger than $\xi\omega_n$.

× Can't be approximated as 2nd-order system.

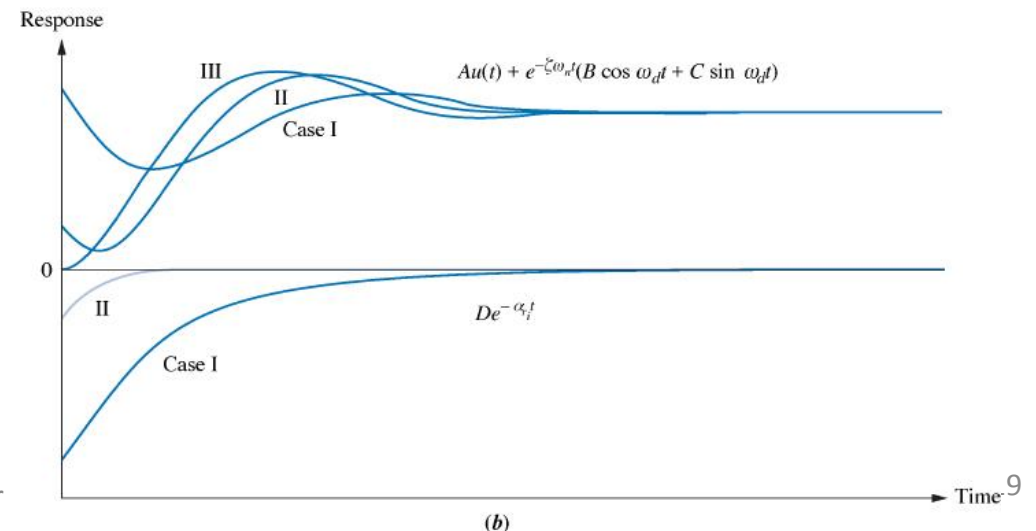


Case II: α_r is much larger than $\xi\omega_n$.

Can be approximated as 2nd-order system.

Case III: $\alpha_r = \infty$.

Pure 2nd-order system.

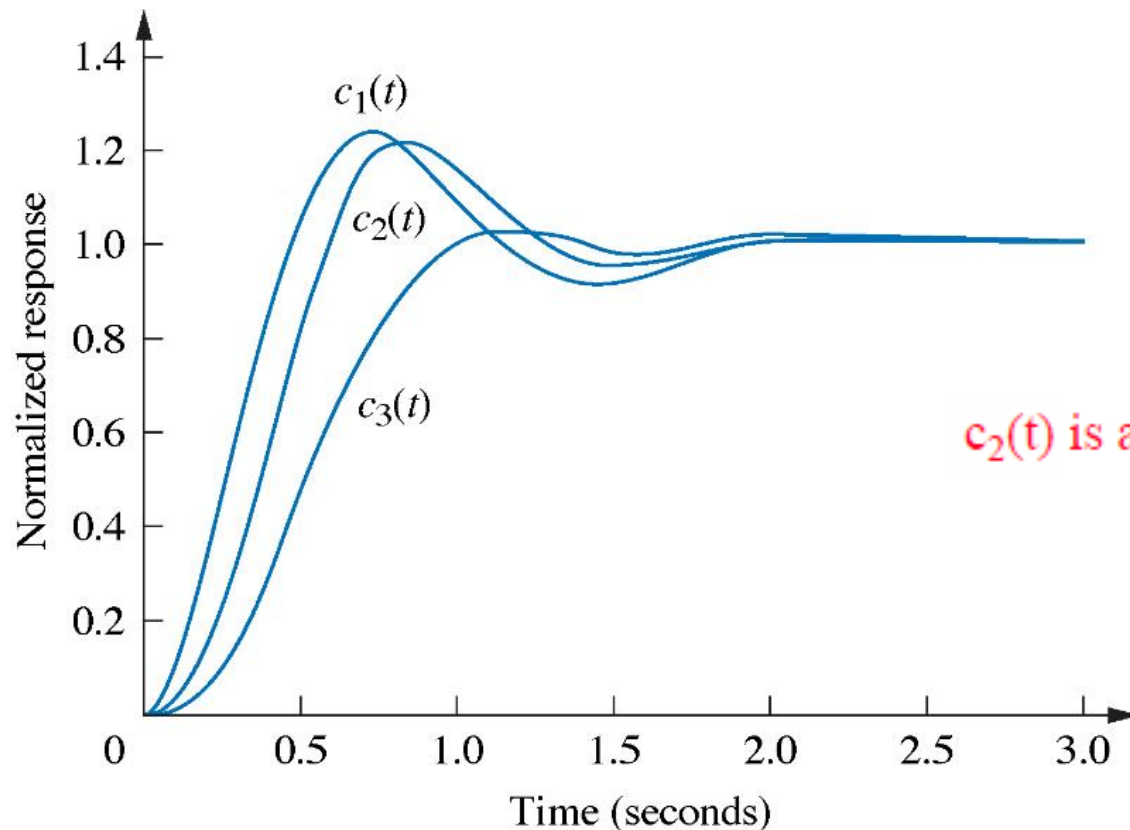


Comparing Responses of Three-Pole Systems

$$T_1(s) = \frac{24.542}{s^2 + 4s + 24.542} \longrightarrow \text{2}^{\text{nd}}\text{-order system}$$

$$T_2(s) = \frac{24.542}{(s+10)(s^2 + 4s + 24.542)} \longrightarrow \text{3}^{\text{rd}}\text{-order system, nondominant pole at -10}$$

$$T_3(s) = \frac{24.542}{(s+3)(s^2 + 4s + 24.542)} \longrightarrow \text{3}^{\text{rd}}\text{-order system, nondominant pole at -3}$$



Evaluating Pole-Zero Cancellation

Problem: For any function for which pole-zero cancellation is valid, find the approximate response.

$$C_1(s) = \frac{26.25(s+4)}{s(s+3.5)(s+5)(s+6)}$$

$$C_2(s) = \frac{26.25(s+4)}{s(s+4.01)(s+5)(s+6)}$$

Zero is at -4

Solution:

The partial-fraction expansion of $C_1(t)$ is

$$C_1(s) = \frac{1}{s} - \frac{3.5}{s+5} + \frac{3.5}{s+6} - \frac{1}{s+3.5}$$

Residue = 1

That residue (1) is not negligible. So a 2nd-order step response approximation cannot be made for $C_1(t)$.

Nearest pole = -3.5 close to zero

The partial-fraction expansion of $C_2(t)$ is

$$C_2(s) = \frac{0.87}{s} - \frac{5.3}{s+5} + \frac{4.4}{s+6} + \frac{0.033}{s+4.01}$$

Residue = 0.033

Nearest pole

That residue (0.033) is negligible, so cancel zero and that pole.

Hence, the approximate response, $c_2(t) = 0.87 - 5.3e^{-5t} + 4.4e^{-6t}$