

Chapter 3

Digital Signals and Systems

Introduction

- In this chapter we discuss the basic concepts and the mathematical tools that form the basis for the representation and analysis of *discrete-time signals* and *systems*.
- The chapter continues to study some properties of *linear systems* such as *time invariance*, *causality*, *impulse response*, *difference equations*, and *digital convolution*.

Digital Signals

- A discrete-time signal $x[n]$ is a sequence of numbers of an integer variable n , where $n \in \mathbb{Z}$.

$x(0)$: zero-th sample amplitude at the sample number $n = 0$,
 $x(1)$: first sample amplitude at the sample number $n = 1$,
 $x(2)$: second sample amplitude at the sample number $n = 2$,
 $x(3)$: third sample amplitude at the sample number $n = 3$, and so on.

- The *duration* or *length* L_x of $x[n]$ is the number of samples from the first nonzero sample $x[n_1]$ to the last nonzero sample $x[n_2]$, $L_x = n_2 - n_1 + 1$.

- the *support* of the sequence is the range $n_1 \leq n \leq n_2$ or $[n_1, n_2]$.

- The symbol \uparrow denotes the index $n = 0$; it is omitted when the table starts at $n = 0$.

Functional

$$x[n] = \begin{cases} \left(\frac{1}{2}\right)^n, & n \geq 0 \\ 0, & n < 0 \end{cases}$$

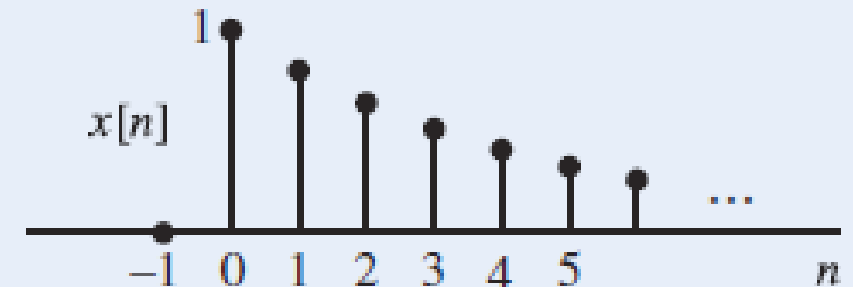
Tabular

n	...	-2	-1	0	1	2	3	...
$x[n]$...	0	0	1	$\frac{1}{2}$	$\frac{1}{4}$	$\frac{1}{8}$...

Sequence

$$x[n] = \{ \dots 0 \underset{\uparrow}{1} \frac{1}{2} \frac{1}{4} \frac{1}{8} \dots \}$$

Pictorial



Signal representation

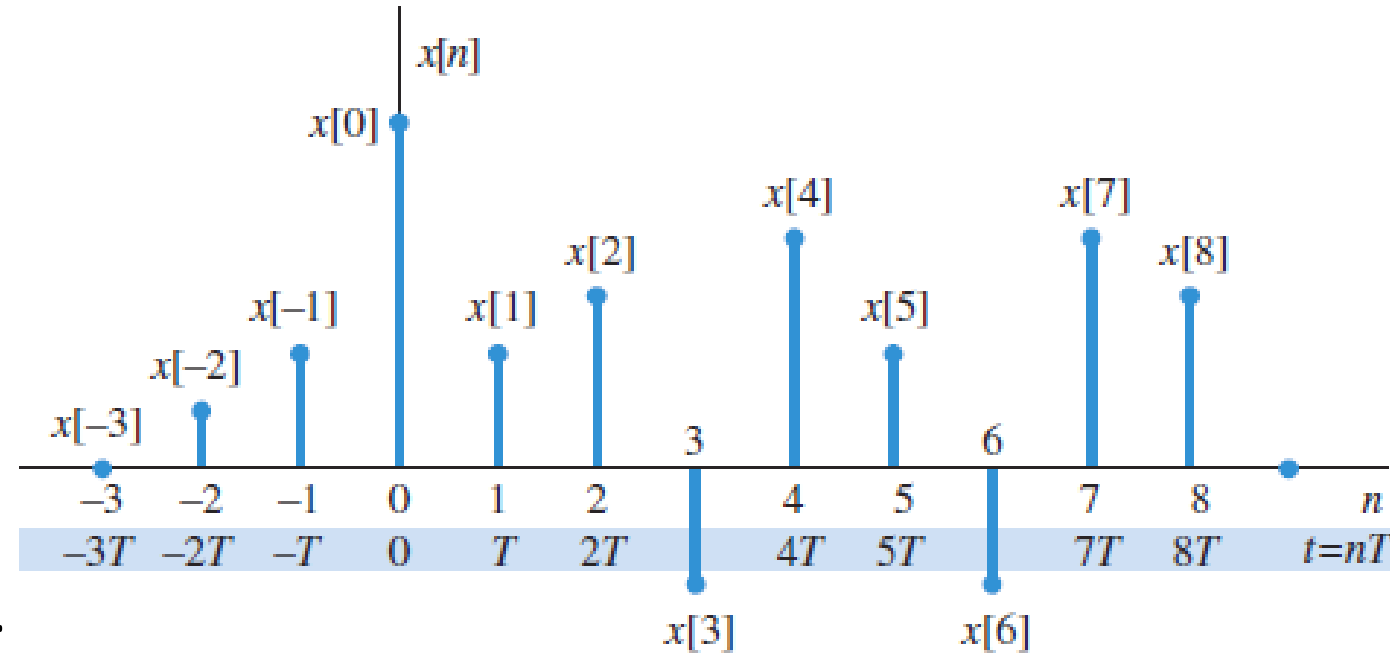
Digital Signals Contd.

Energy: The energy of a sequence $x[n]$ is defined by the formula

$$\mathcal{E}_x \triangleq \sum_{n=-\infty}^{\infty} |x[n]|^2.$$

Power: the power of a sequence $x[n]$ is defined as the average energy per sample.

$$\mathcal{P}_x \triangleq \lim_{L \rightarrow \infty} \left[\frac{1}{2L+1} \sum_{n=-L}^L |x[n]|^2 \right].$$

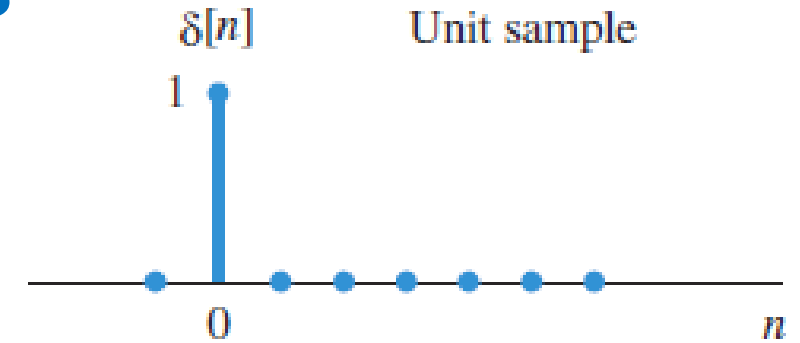


Representation of a sampled signal

Common Digital Sequences

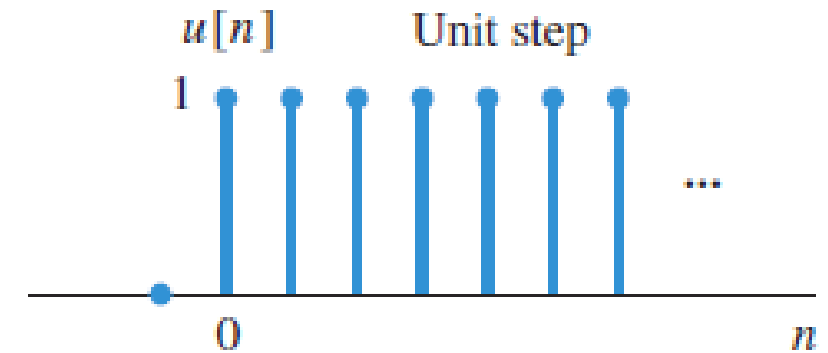
1. Unit-impulse sequence:

$$\delta[n] \triangleq \begin{cases} 1, & n = 0 \\ 0, & n \neq 0 \end{cases}$$



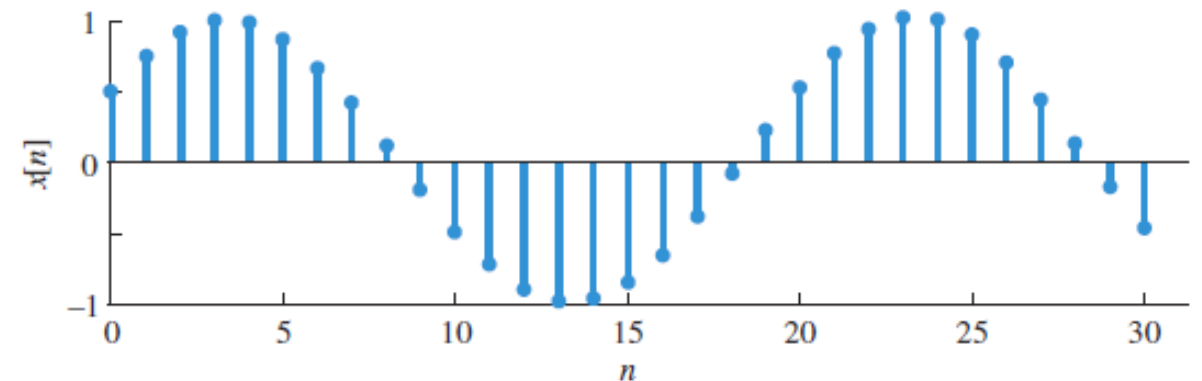
2. Unit-step sequence:

$$u[n] \triangleq \begin{cases} 1, & n \geq 0 \\ 0, & n < 0 \end{cases}$$



3. Sinusoidal sequence

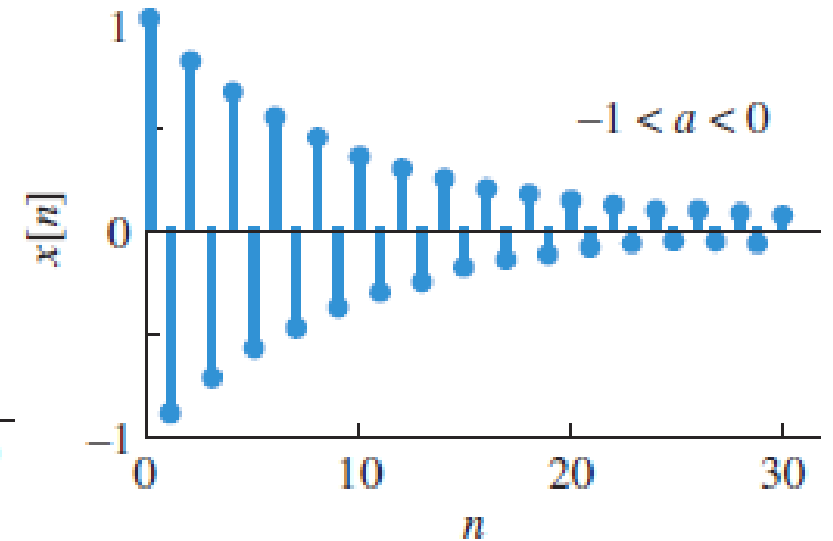
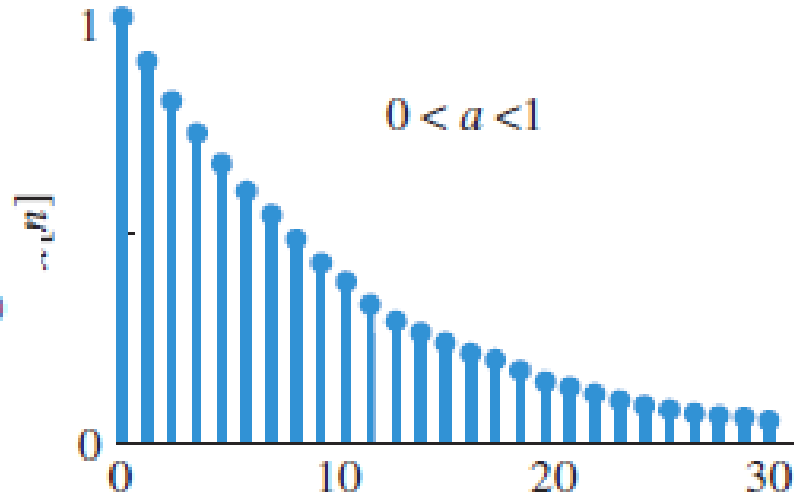
$$x[n] = A \cos(\omega_0 n + \phi), \quad -\infty < n < \infty$$



Common Digital Sequences

4. Exponential sequence

$$x[n] \triangleq Aa^n, \quad -\infty < n < \infty$$



where A and a can take real or complex values. n

Signal generation and plotting in MATLAB

```
n=(-10:10); x=2*cos(2*pi*0.05*n);
```

generate a unit sample and unit step sequences at $n=n_0$ in the range $n=(n_1:n_2)$.

```
[x,n] = delta(n1,n0,n2); % Unit impulse sequence
```

```
[x,n] = unitstep(n1,n0,n2); % Unit step sequence
```

```
[x,n] = unitpulse(n1,n2,n3,n4); % Unit pulse sequence
```

creates a rectangular pulse of unit amplitude from n_2 to n_3 and zero elsewhere

```
x = perseggen(xp,Np,Nps); % Periodic sequence
```

generates Nps periods of a periodic signal with period Np .

```
stem(n,x,'fill'); ylabel('x[n]'); xlabel('n'); plot the sequence as a discrete-time signal
```

Operations on sequences

- $y[n] = x_1[n] + x_2[n]$, (signal addition)
- $y[n] = x_1[n] - x_2[n]$, (signal subtraction)
- $y[n] = x_1[n] \cdot x_2[n]$, (signal multiplication)
- $y[n] = x_1[n]/x_2[n]$, (signal division)
- $y[n] = a \cdot x_2[n]$, (signal scaling)

- Time-reversal or folding:** reflects the sequence $x[n]$ about the origin $n = 0$. $y[n] = x[-n]$,

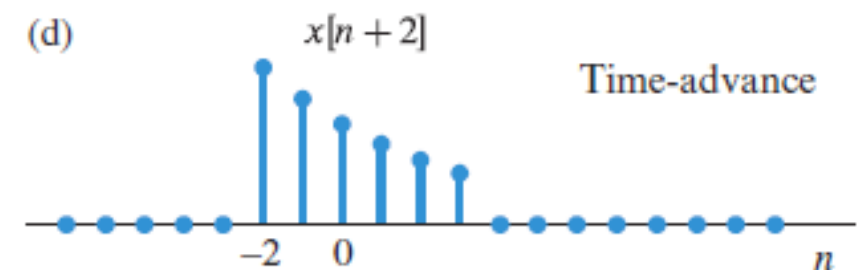
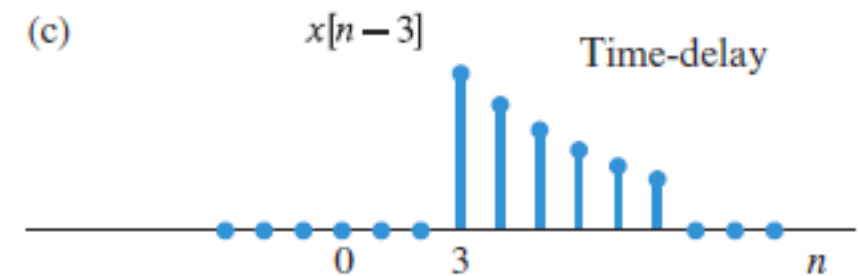
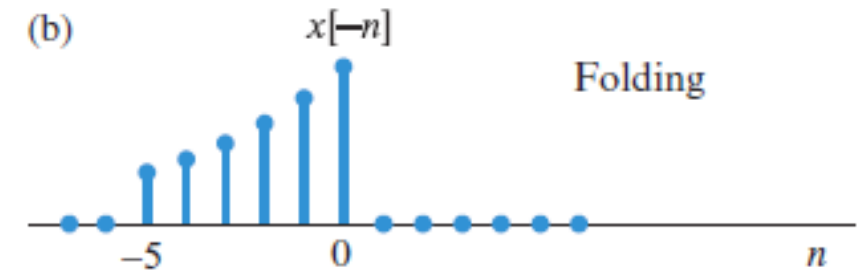
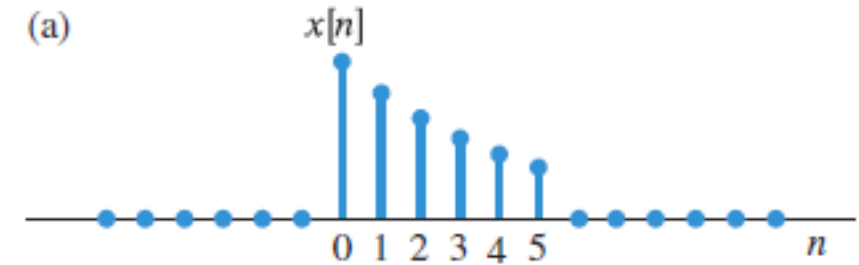
$$x[-n] = x[n] \quad \text{even or symmetric}$$

$$x[-n] = -x[n] \quad \text{odd or anti-symmetric}$$

- Time-shifting:** the sequence $x[n]$ is shifted by n_0 samples
 $y[n] = x[n - n_0]$.

If $n_0 > 0$ *shift to the right (time-delay)*

If $n_0 < 0$ *shift to the left (time-advance.)*



Folding and time-shifting

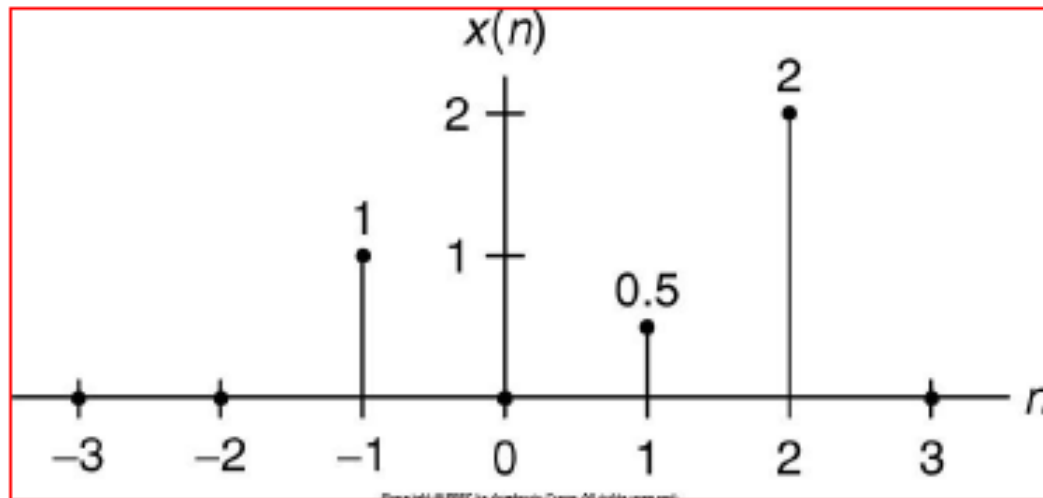
Example 1

Given the following,

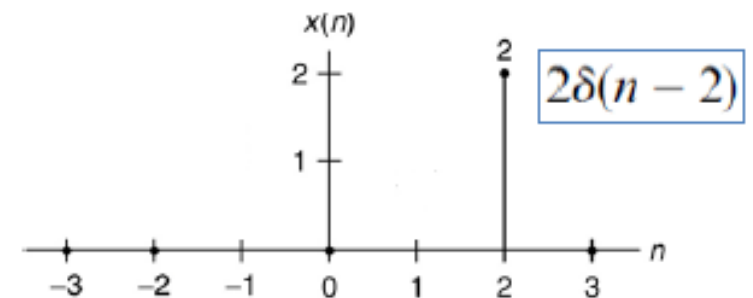
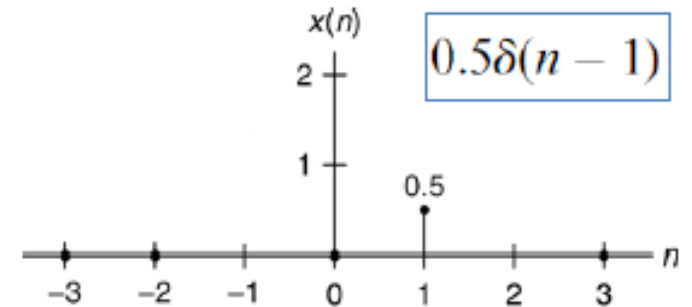
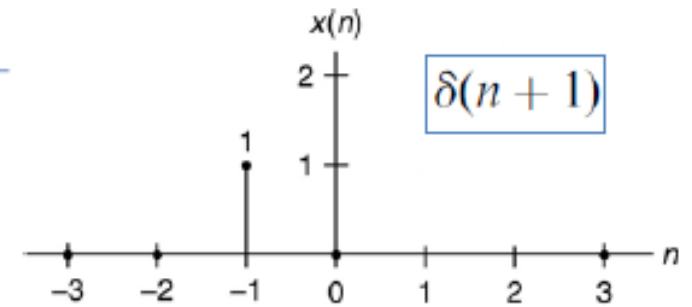
$$x(n) = \delta(n + 1) + 0.5\delta(n - 1) + 2\delta(n - 2),$$

a. Sketch this sequence.

Solution:



The sum



Generation of Digital Signals

- To generate the digital sequence $x(n)$ from the analog signal $x(t)$:

uniformly sampling at the time interval of $\Delta t = T$

$$x(n) = x(t) \Big|_{t=nT} = x(nT)$$

Example 2

Convert analog signal $x(t)$ into digital signal $x(n)$, when sampling period is 125 microsecond, also plot sample values.

$$x(t) = 10e^{-5000t}u(t)$$

Solution:

$$t = nT = n \times 0.000125 = 0.000125n$$

$$x(n) = x(nT) = 10e^{-5000 \times 0.000125n}u(nT) = 10e^{-0.625n}u(n)$$

Example 2 (contd.)

The first five
sample values:



$$x(0) = 10e^{-0.625 \times 0} u(0) = 10.0$$

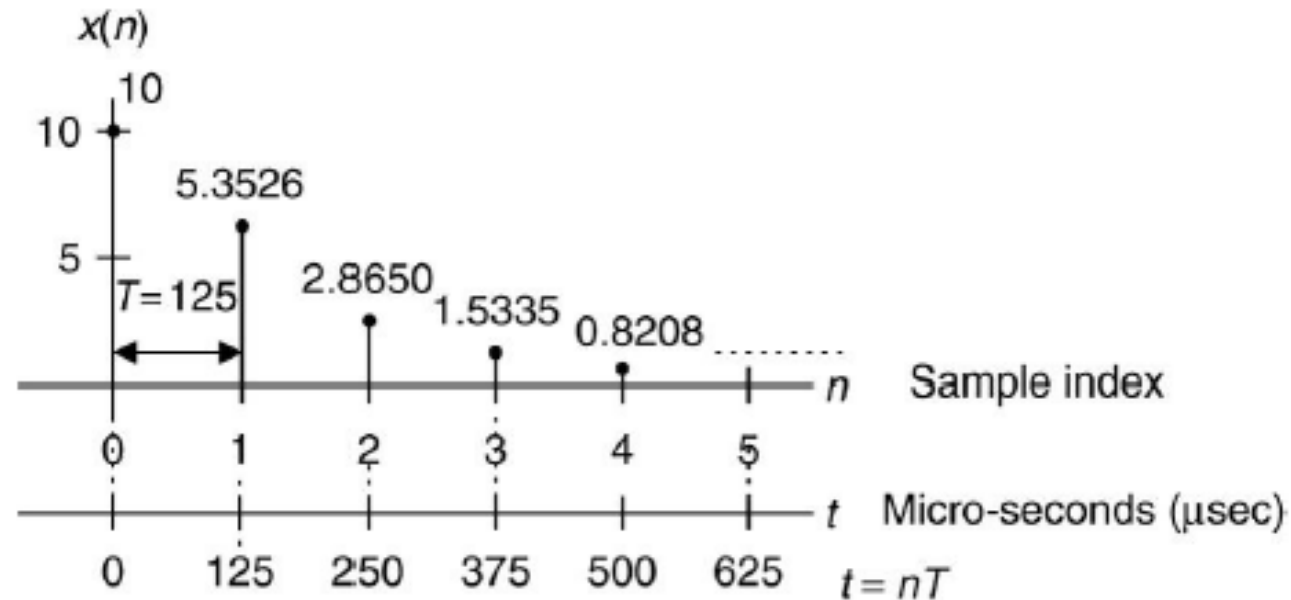
$$x(1) = 10e^{-0.625 \times 1} u(1) = 5.3526$$

$$x(2) = 10e^{-0.625 \times 2} u(2) = 2.8650$$

$$x(3) = 10e^{-0.625 \times 3} u(3) = 1.5335$$

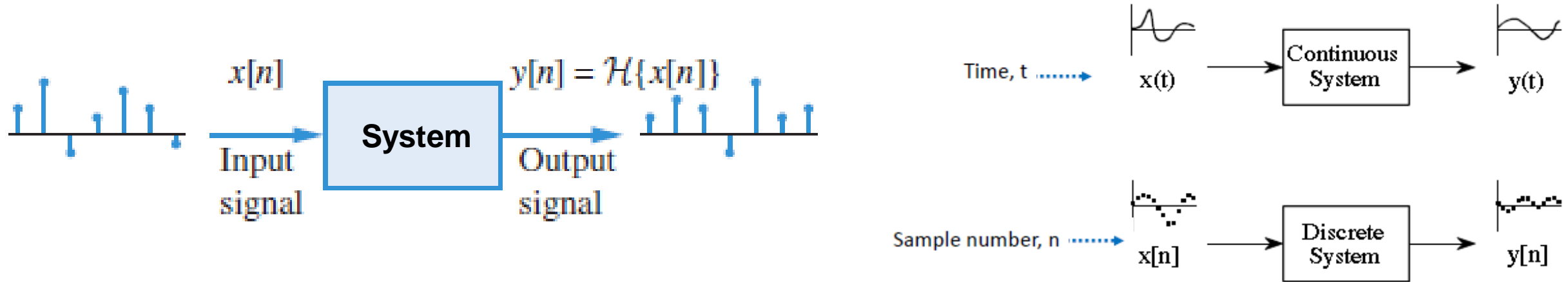
$$x(4) = 10e^{-0.625 \times 4} u(4) = 0.8208$$

Plot of the digital
sequence:



Digital Systems

A **Digital system** is a computational process or algorithm that transforms or maps a sequence $x[n]$, called the **input signal**, into another sequence $y[n]$, called the **output signal**.



Example 3

Determine the response of the following system to the input signal $x(n) = \begin{cases} |n|, & -3 \leq n \leq 3 \\ 0, & \text{otherwise} \end{cases}$ and the system's output $y(n) = \frac{1}{3}[x(n+1) + x(n) + x(n-1)]$

Solution:

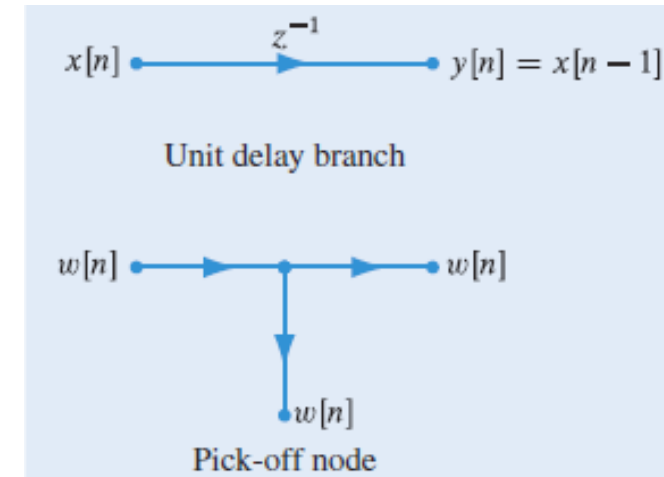
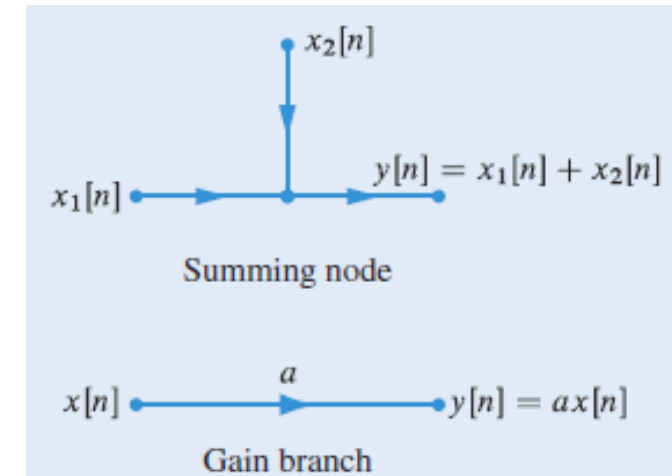
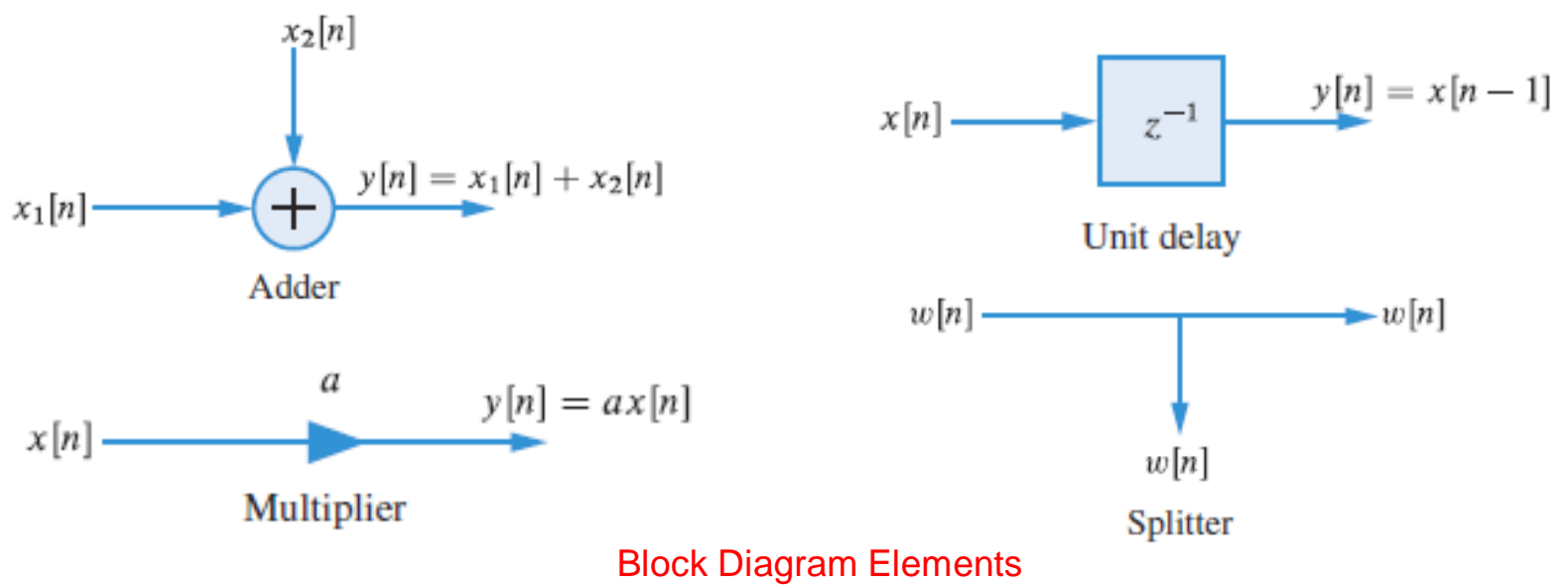
The output of this system is the mean value of the present, immediate past, and the immediate future samples.

$$\text{For } n = 0 \Rightarrow y(0) = \frac{1}{3}[x(-1) + x(0) + x(1)] = \frac{1}{3}[1 + 0 + 1] = \frac{2}{3}$$

$$\text{Repeating this computation for every value of } n \Rightarrow y(n) = \left\{ \dots, 0, 1, \frac{5}{3}, 2, 1, \frac{2}{3}, 1, 2, \frac{5}{3}, 1, 0, \dots \right\}$$

Block Diagram of Discrete-Time Systems

- Operations required in the implementation of a discrete-time system can be depicted in one of two ways: a *block diagram* or a *signal flow graph*.

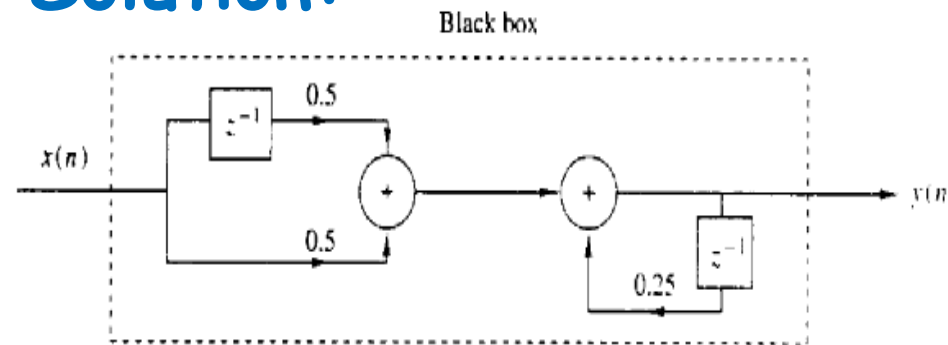


Example 4

Sketch the block diagram representation of the discrete-time system described by the input-output relation.

$$y(n) = \frac{1}{4}y(n-1) + \frac{1}{2}x(n) + \frac{1}{2}x(n-1)$$

Solution:



Classification of Discrete-Time Systems

Static and dynamic systems

$y(n) = x(n) + 3x(n-1)$ Dynamic (have memory) past or future samples of the input

$y(n) = n x(n) + b x^3(n)$ Static or memory-less no past or future samples of the input

Causality

A system is called **causal** if the present value of the output does not depend on future values of the input. Causality is required for systems that should operate in real-time.

Example: $y(n) = 0.5x(n) + 2.5x(n-2)$, for $n \geq 0$

- If the output of a system depends on future values of its input, the system is **non-causal**.

Example: $y(n) = 0.25x(n-1) + 0.5x(n+1) - 0.4y(n-1)$, for $n \geq 0$

Stability

A system is said to be **stable**, in the Bounded-Input Bounded-Output (BIBO) sense. Stability is a property that should be satisfied by every practical system.

Example: $|x[n]| \leq M_x < \infty \Rightarrow |y[n]| \leq M_y < \infty$.

- The moving-average system is **stable**: $y[n] = \frac{1}{3}[x[n] + x[n-1] + x[n-2]] \leq 3\bar{M}_x$ for $|x[n]| \leq M_x$
- The accumulator system is **unstable**: $y[n] = \sum_{k=0}^{\infty} x[n-k]$ becomes unbounded as $n \rightarrow \infty$.

Example 9

Given a linear system given by: $y(n) = 0.25y(n-1) + x(n)$ for $n \geq 0$ and $y(-1) = 0$

Which is described by the unit-impulse response: $h(n) = (0.25)^n u(n)$

Determine whether the system is stable or not.

Solution: To determine whether a system is stable, we apply the following equation:

$$S = \sum_{k=-\infty}^{\infty} |h(k)| = \dots + |h(-1)| + |h(0)| + |h(1)| + \dots < \infty.$$

$$S = \sum_{k=-\infty}^{\infty} |h(k)| = \sum_{k=-\infty}^{\infty} |(0.25)^k u(k)|$$

Using definition of step function: $u(k) = 1$ for $k \geq 0$, $\Rightarrow S = \sum_{k=0}^{\infty} (0.25)^k = 1 + 0.25 + 0.25^2 + \dots$

For $a < 1$, we know $\sum_{k=0}^{\infty} a^k = \frac{1}{1-a}$ where $a = 0.25 < 1$

Therefore $S = 1 + 0.25 + 0.25^2 + \dots = \frac{1}{1-0.25} = \frac{4}{3} < \infty$

**The summation is finite,
so the system is stable.**

Linearity

A digital system is *linear* if and only if it satisfy the *superposition principle*, for every real or complex constant a_1, a_2 and every input signal $x_1[n]$ and $x_2[n]$:

$$H[a_1x_1[n] + a_2x_2[n]] = a_1H[x_1[n]] + a_2H[x_2[n]] \quad \leftarrow \text{Homogeneity \& Additivity}$$

Homogeneity
(deals with amplitude)

Additivity

linearity means that the output due to a sum of input signals equals the sum of outputs due to each signal alone

time invariance

system is called *time-invariant* or fixed if and only if for every input $x[n]$ and every time shift n_0

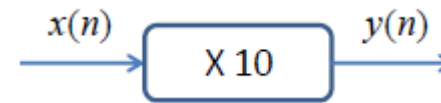
$$y[n] = \mathcal{H}\{x[n]\} \Rightarrow y[n - n_0] = \mathcal{H}\{x[n - n_0]\}, \quad \leftarrow \text{a time shift in the input results in a corresponding time shift in the output}$$

Time-invariance means that the system does not change over time.

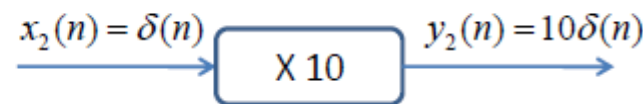
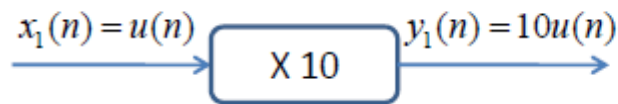
Example 5 (Linear System)

Let a digital amplifier, $y(n) = 10x(n)$

If the inputs are: $x_1(n) = u(n)$ and $x_2(n) = \delta(n)$



Outputs will be: $y_1(n) = 10u(n)$ and $y_2(n) = 10\delta(n)$, respectively.



If we apply combined input to the system: $x(n) = 2x_1(n) + 4x_2(n) = 2u(n) + 4\delta(n)$

The output will be: $y(n) = 10x(n) = 10(2u(n) + 4\delta(n)) = 20u(n) + 40\delta(n)$

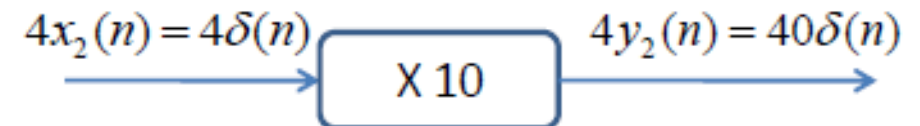
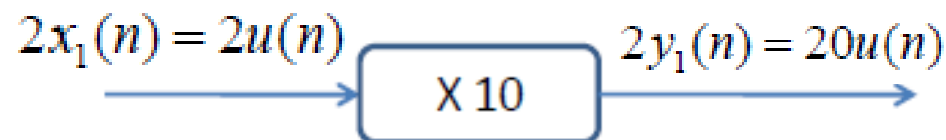
Linear
System

Individual outputs:

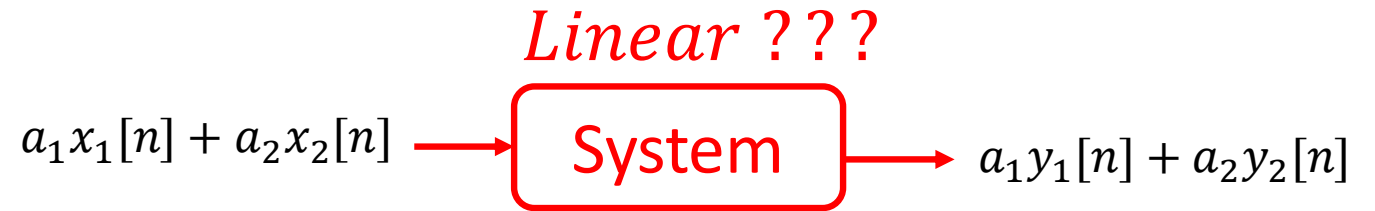
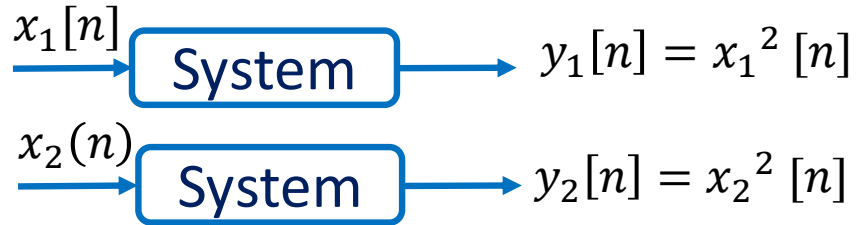
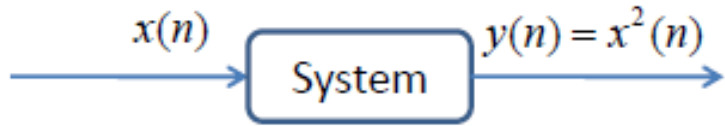
$$2y_1(n) = 2 \times 10x_1(n) = 20u(n)$$

$$4y_2(n) = 4 \times 10x_2(n) = 40\delta(n)$$

$$y(n) = 2y_1(n) + 4y_2(n)$$



Example 5 (Non-Linear System)



If the input is: $x[n] = a_1x_1[n] + a_2x_2[n]$

Then the output is: $y[n] = (x[n])^2 = (a_1x_1[n] + a_2x_2[n])^2 = a_1^2x_1^2[n] + a_2^2x_2^2[n] + 2a_2a_1x_1[n]x_2[n]$ (1)

Individual outputs: $a_1y_1[n] = a_1x_1^2[n]$
 $a_2y_2[n] = a_2x_2^2[n]$ $\Rightarrow a_1y_1[n] + a_2y_2[n] = a_1x_1^2[n] + a_2x_2^2[n]$ (2)

(1) (2)



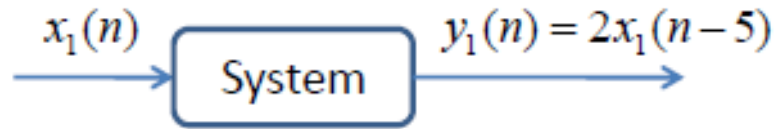
$$y[n] \neq a_1y_1[n] + a_2y_2[n]$$

Non Linear
System

Example 6 (Time Invariant System)

Given the linear system $y[n] = 2x[n - 5]$, find whether the system is time invariant or not.

Solution:



Let the shifted input be: $x_2[n] = x_1[n - n_0]$,

Therefore system output: $y_2[n] = 2x_2[n - 5] = 2x_1[n - n_0 - 5]$,

Shifting $y_1[n] = 2x_1[n - 5]$ by n_0 samples leads to $y_1[n - n_0] = 2x_1[n - 5 - n_0]$

Equal

Time Invariant

Example 6 (Not Time Invariant System)

Given the linear system $y(n) = n x(n)$, find whether the system is time invariant or not.

Solution:



Let the shifted input be: $x_2(n) = x_1(n - n_0)$

Therefore system output due to the shifted input $\Rightarrow y_2(n) = nx_2(n) = nx_1(n - n_0)$

Shifting $y_1(n) = nx_1(n)$ by n_0 samples (replace n by $n - n_0$) leads to

$$y_1(n - n_0) = (n - n_0)x_1(n - n_0)$$

NOT Equal

NOT Time Invariant

Difference Equation

A causal, linear, time-invariant system (LTI) can be described by a ***difference equation*** as follow:

$$y(n) + a_1 y(n-1) + \cdots + a_N y(n-N) = b_0 x(n) + b_1 x(n-1) + \cdots + b_M x(n-M)$$

Outputs

Inputs

After rearranging: $y(n) = -a_1 y(n-1) - \cdots - a_N y(n-N) + b_0 x(n) + b_1 x(n-1) + \cdots + b_M x(n-M)$

Finally:

$$y(n) = -\sum_{i=1}^N a_i y(n-i) + \sum_{j=0}^M b_j x(n-j)$$

Example 7 Identify non zero system coefficients of the following difference equations.

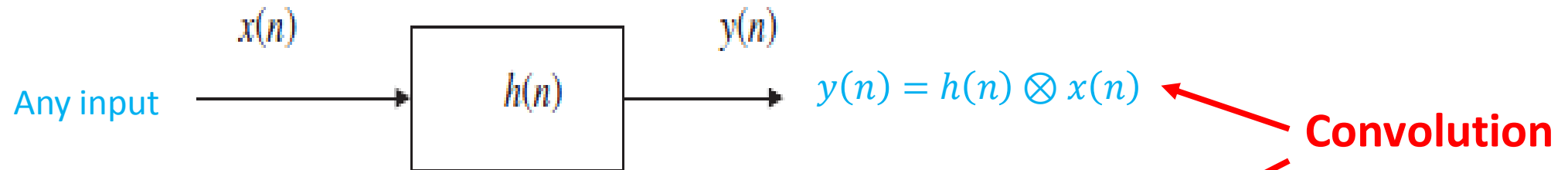
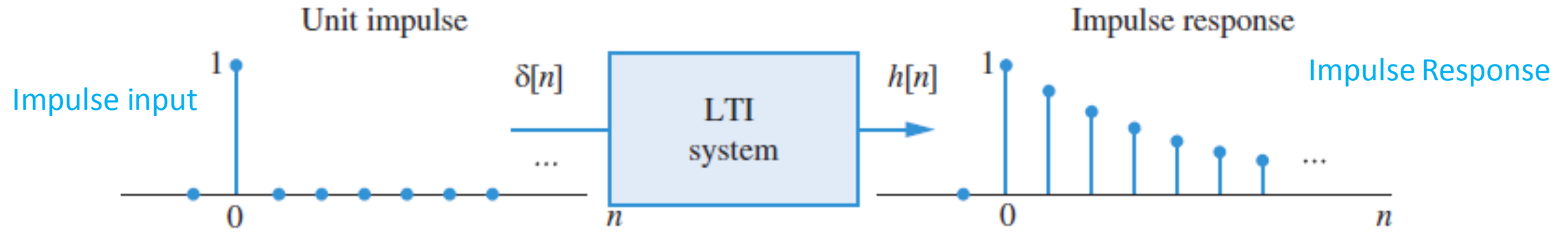
Solution:

$$y(n) = 0.25y(n-1) + x(n) \longrightarrow b_0 = 1, \quad a_1 = -0.25$$

$$y(n) = x(n) + 0.5x(n-1) \longrightarrow b_0 = 1, \quad b_1 = 0.5$$

System Representation Using Impulse Response

- The *linearity* and *time invariance* properties greatly simplify the analysis of linear systems (output of a decomposed, scaled and shifted input signal = sum of outputs of individual inputs)
- A *linear time-invariant* system (LTI system) can be completely described by its *impulse response* $h[n]$ due to the impulse input $\delta(n)$ with zero initial conditions.



$$y(n) = \sum_{k=-\infty}^{\infty} h(k)x(n-k) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$

$$y(n) = \cdots + h(-1)x(n+1) + h(0)x(n) + h(1)x(n-1) + h(2)x(n-2) + \cdots$$

Example 8 (a)

Given the linear time-invariant system:

$$y(n) = 0.5x(n) + 0.25x(n-1) \text{ with an initial condition } x(-1) = 0,$$

- Determine the unit-impulse response $h(n)$.
- Draw the system block diagram.
- Write the output using the obtained impulse response.

Solution:

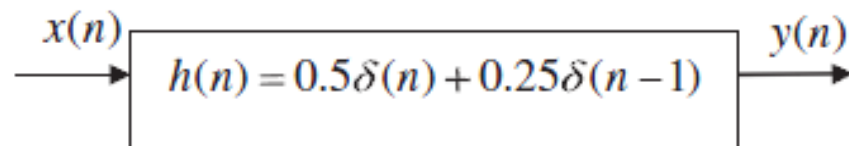
a. let $x(n) = \delta(n)$, then $h(n) = y(n) = 0.5x(n) + 0.25x(n-1) = 0.5\delta(n) + 0.25\delta(n-1)$

Therefore,
$$h(n) = \begin{cases} 0.5 & n = 0 \\ 0.25 & n = 1 \\ 0 & \text{elsewhere} \end{cases}$$

c. The system output

$$y(n) = h(0)x(n) + h(1)x(n-1)$$

b. The block diagram of the LTI system



Example 8 (b)

Given the difference equation

$$y(n] = 0.25y(n - 1) + x(n) \text{ for } n \geq 0 \text{ and } y(-1) = 0,$$

- Determine the unit-impulse response $h(n)$.
- Draw the system block diagram.
- Write the output using the obtained impulse response.

Solution:

a. Let $x(n) = \delta(n)$, then $h(n) = 0.25h(n - 1) + \delta(n)$

To solve for $h(n)$, we evaluate

$$h(0) = 0.25h(-1) + \delta(0) = 0.25 \times 0 + 1 = 1$$

$$h(1) = 0.25h(0) + \delta(1) = 0.25 \times 1 + 0 = 0.25$$

$$h(2) = 0.25h(1) + \delta(2) = 0.25 \times 0.25 + 0 = 0.0625$$

....

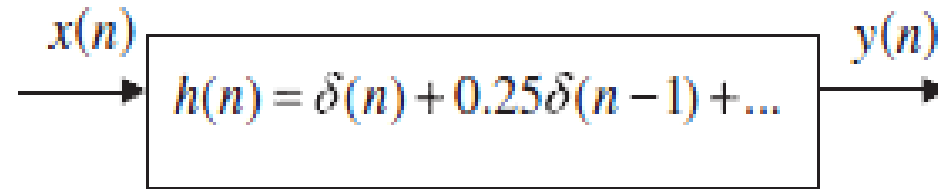
With the calculated results, we can predict the impulse response as

$$h(n) = (0.25)^n u(n) = \delta(n) + 0.25\delta(n - 1) + 0.0625\delta(n - 2) + \dots$$

 Infinite!

Example 8 (b) – contd.

b. The system block diagram



c. The output sequence

$$\begin{aligned} y(n) &= h(0)x(n) + h(1)x(n-1) + h(2)x(n-2) + \dots \\ &= x(n) + 0.25x(n-1) + 0.0625x(n-2) + \dots \end{aligned}$$

Finite Impulse Response (FIR) system:

When the difference equation contains no previous outputs, i.e. ' a ' coefficients are zero.
(See example 8 (a))

Infinite Impulse Response (IIR) system:

When the difference equation contains previous outputs, i.e. ' a ' coefficients are not all zero. (See example 8 (b))

Digital Convolution

- A LTI system can be represented using a digital convolution

$$y(n) = h(n) * x(n) \Rightarrow y(n) = \sum_{k=-\infty}^{\infty} h(k)x(n-k) \\ = \cdots + h(-1)x(n+1) + h(0)x(n) + h(1)x(n-1) + h(2)x(n-2) + \cdots$$

- The unit-impulse response $h(n)$ relates the system input and output.

- The sequences are interchangeable.

$$y(n) = \sum_{k=-\infty}^{\infty} h(k)x(n-k) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$

Commutative

$x[n] * h[n] = h[n] * x[n]$

- Convolution sum requires $h(n)$ to be reversed and shifted.
- If $h(n)$ is the given sequence, $h(-n)$ is the ***reversed sequence***.

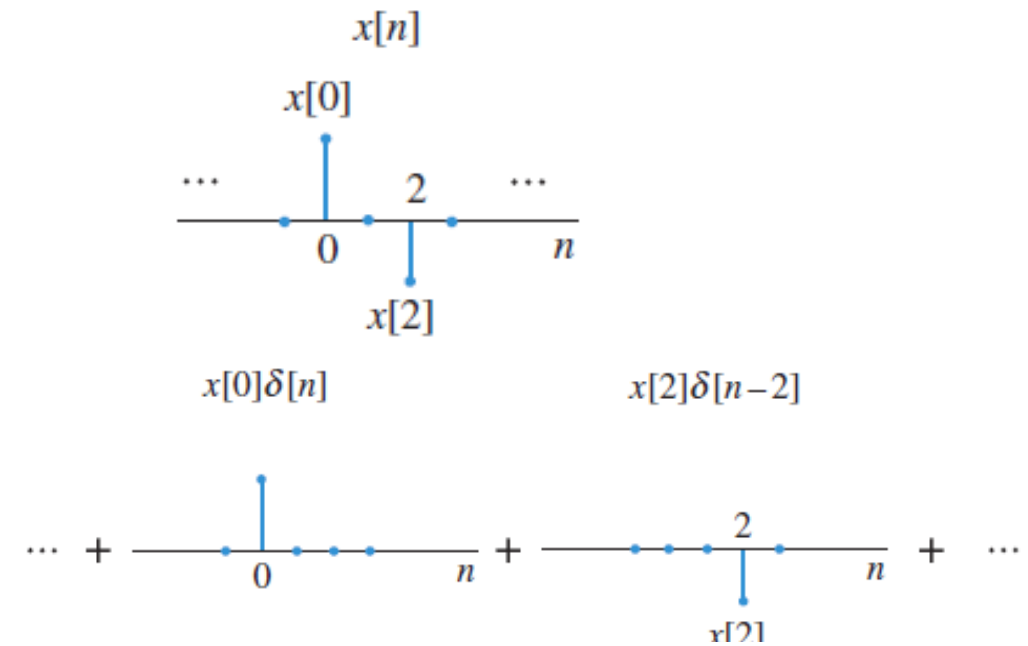
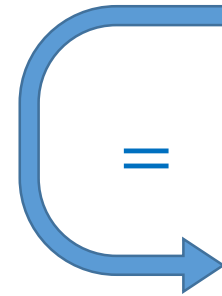
Signal decomposition into impulses

$$x_k[n] = \begin{cases} x[k], & n = k \\ 0, & n \neq k \end{cases} \leftarrow \text{Sample } k \text{ of } x[n]$$

$$x_k[n] = \delta[n - k]; \leftarrow \text{Impulse at } n = k$$

$$\delta[n - k] = \begin{cases} 1, & n = k \\ 0, & n \neq k \end{cases} \leftarrow \text{Shifted Impulse}$$

$$x[n] = \sum_{k=-\infty}^{\infty} x[k] \delta[n - k]. \quad -\infty < n < \infty$$



Reversed Sequence

Example: Given a sequence $h(k) = \begin{cases} 3, & k = 0, 1 \\ 1, & k = 2, 3 \\ 0 & \text{elsewhere} \end{cases}$

Sketch the sequence $h(k)$ and reversed sequence $h(-k)$.

Solution:

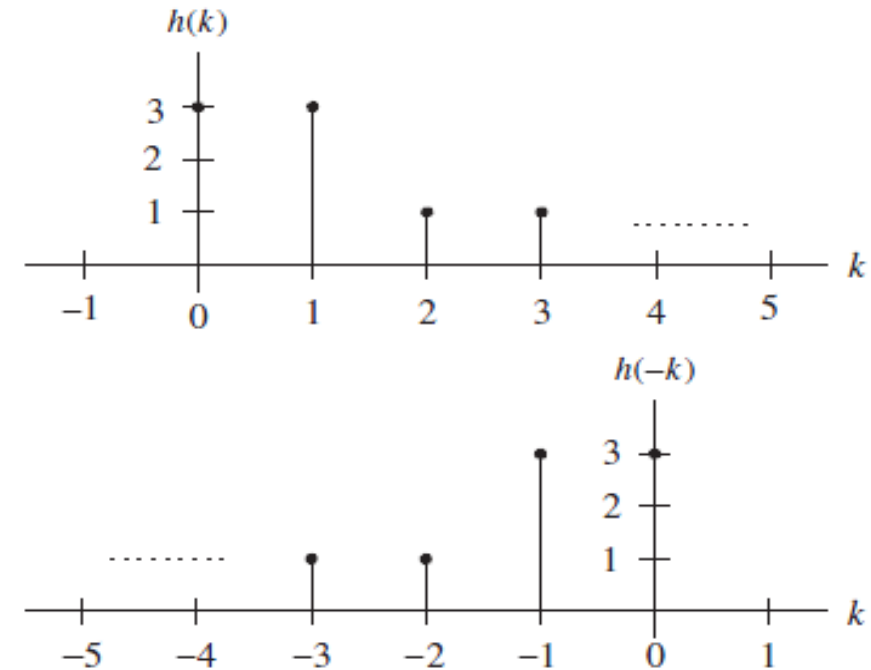
$$k > 0, h(-k) = 0$$

$$k = 0, h(-0) = h(0) = 3$$

$$k = -1, h(-k) = h(-(-1)) = h(1) = 3$$

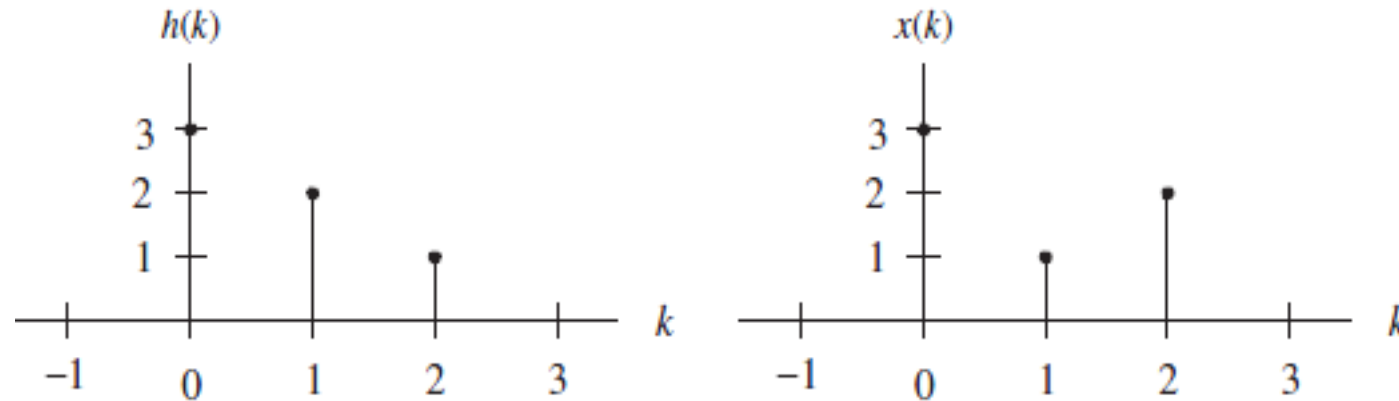
$$k = -2, h(-k) = h(-(-2)) = h(2) = 1$$

$$k = -3, h(-k) = h(-(-3)) = h(3) = 1$$



Convolution Using Table Method

Example 9



Solution:

Length = 3

Length = 3

Convolution sum using the table method.

k :	-2	-1	0	1	2	3	4	5	
$x(k)$:			3	1	2				
$h(-k)$:	1	2	3						$y(0) = 3 \times 3 = 9$
$h(1-k)$:		1	2	3					$y(1) = 3 \times 2 + 1 \times 3 = 9$
$h(2-k)$:			1	2	3				$y(2) = 3 \times 1 + 1 \times 2 + 2 \times 3 = 11$
$h(3-k)$:				1	2	3			$y(3) = 1 \times 1 + 2 \times 2 = 5$
$h(4-k)$:					1	2	3		$y(4) = 2 \times 1 = 2$
$h(5-k)$:						1	2	3	$y(5) = 0$ (no overlap)

Convolution Using Table Method

Example 10

$$x(n) = \begin{cases} 1 & n = 0, 1, 2 \\ 0 & \text{otherwise} \end{cases} \text{ and } h(n) = \begin{cases} 0 & n = 0 \\ 1 & n = 1, 2 \\ 0 & \text{otherwise} \end{cases}$$

Solution:

Length = 3

Length = 2

$k:$	-2	-1	0	1	2	3	4	5	...
$x(k):$			1	1	1				...
$h(-k):$	1	1	0						$y(0) = 0$ (no overlap)
$h(1-k):$		1	1	0					$y(1) = 1 \times 1 = 1$
$h(2-k):$			1	1	0				$y(2) = 1 \times 1 + 1 \times 1 = 2$
$h(3-k):$				1	1	0			$y(3) = 1 \times 1 + 1 \times 1 = 2$
$h(4-k):$					1	1	0		$y(4) = 1 \times 1 = 1$
$h(n-k):$						1	1	0	$y(n) = 0, n \geq 5$ (no overlap)
									Stop

Convolution length = 3 + 2 - 1 = 4

Convolution Properties

- $\delta[n]$ is the *identity element* of the convolution operation.
- Commutative: $a[n] \otimes b[n] = b[n] \otimes a[n]$
- Associative: $(a[n] \otimes b[n]) \otimes c[n] = a[n] \otimes (b[n] \otimes c[n])$
- Distributive: $a[n] \otimes (b[n] + c[n]) = a[n] \otimes b[n] + a[n] \otimes c[n]$

